

A New Boundary Element Method for Multiple-Frequency Parameter Extraction of Lossy Substrates

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Outline

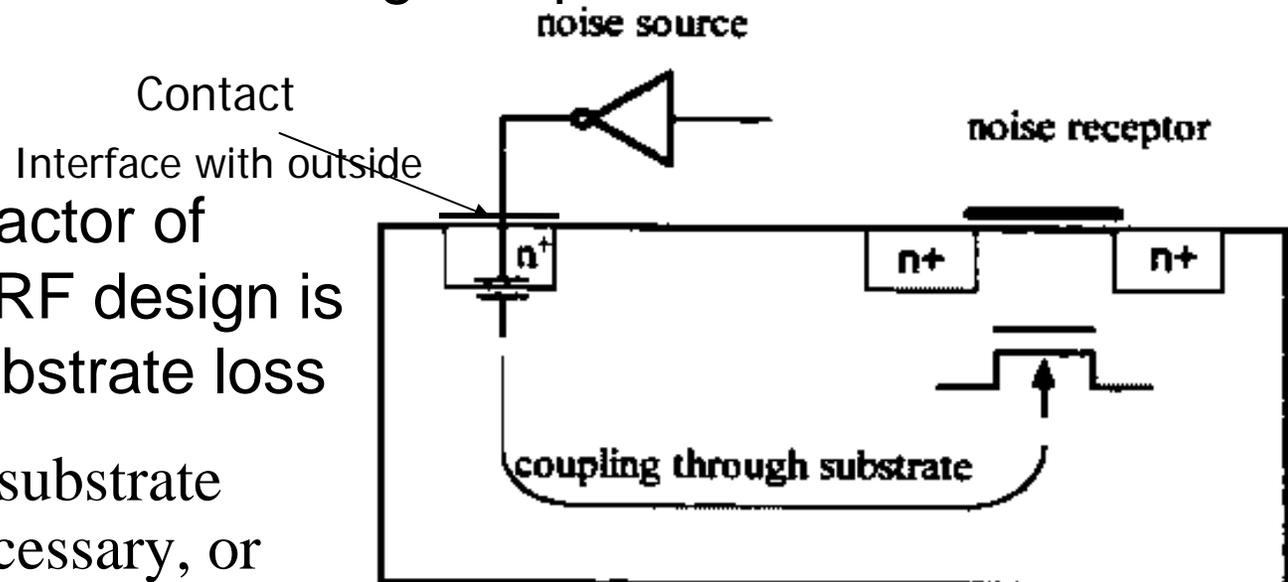
- Introduction of Substrate Coupling Problems
- High-Frequency Parameter Extraction Using Direct Boundary Element Method (DBEM)
- Efficient Techniques for Multi-Frequency Extraction
- Numerical Results
- Conclusions

Introduction of Substrate Coupling Problems

- In mixed-signal circuits
 - Digital and analog components are often built on a single lossy substrate (Si)
 - Coupling noises traveling through the substrate severely impact the sensitive analog components

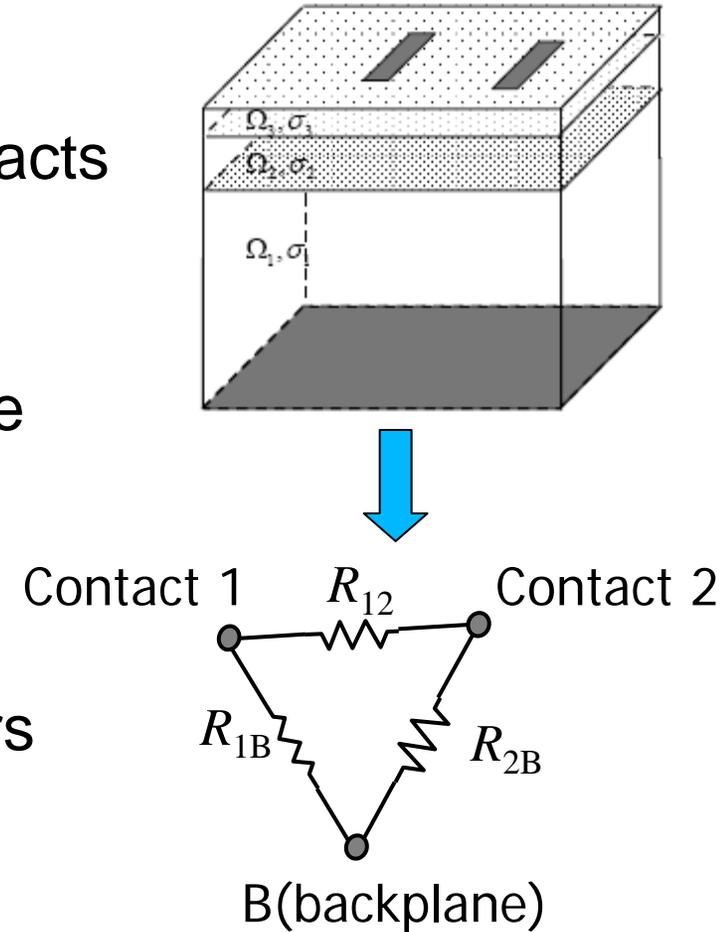
- The quality factor of inductors in RF design is limited by substrate loss

knowledge of substrate coupling is necessary, or even critical for design



Introduction of Substrate Coupling Problems

- Modeling of substrate coupling
 - Substrate resistance among contacts
 - At higher frequency, both resistive and capacitive couplings should be considered [TCAD'98]
 - Extraction of substrate parameters becomes a challenging task



A.M. Niknejad, R. Gharpurey, and R.G. Meyer, "Numerically stable Green function for modeling and analysis of substrate coupling in integrated circuits," *IEEE Trans. CAD*, Vol. 17, pp. 305-315, 1998.

Numerical Methods for Substrate Extraction

- Volume discretization methods
 - Finite Element Method (FEM)
 - Finite Difference Method (FDM)
- Advantage:
 - Versatile for various kinds of substrate structures
 - Stratified, with multiple parallel horizontal layers
 - More complicated, e.g. those with oxide wells, trenches, sinkers, buried diffusions, shielding for noise reduction
- Disadvantage:
 - Too many unknowns
 - Then, limited to small structures

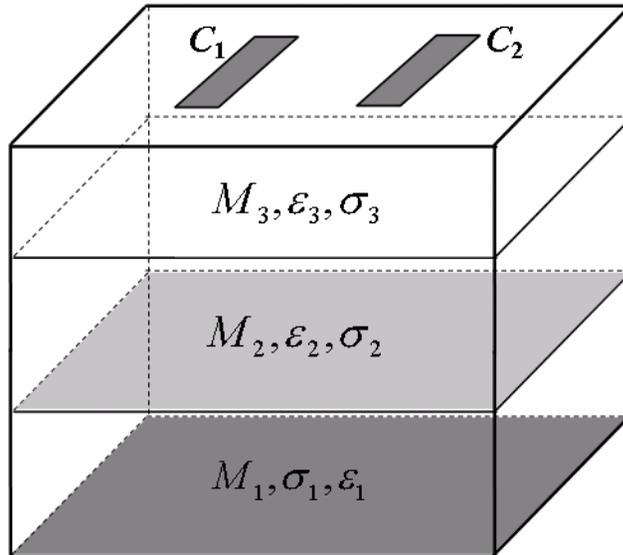
Numerical Methods for Substrate Extraction

- Green's function based methods (BEM)
 - Widely investigated
- Advantages:
 - Only discretizes contact surfaces, involves the fewest variables
 - Acceleration techniques proposed, such as DCT (discrete cosine), eigen-decomposition, etc.
- Disadvantages:
 - The derivation of Green's function involves infinite series, which is expensive to calculate
 - Difficult, or even impossible to find the Green's function for non-stratified substrates

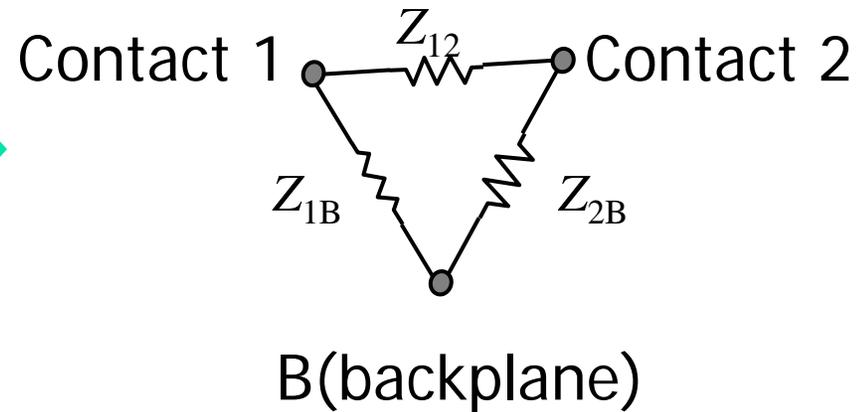
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- Introduction of Substrate Coupling Problems
- **High-Frequency Parameter Extraction Using DBEM**
- Efficient Techniques for Multi-Frequency Extraction
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High-Frequency Model of Substrate



Reduced circuit model:

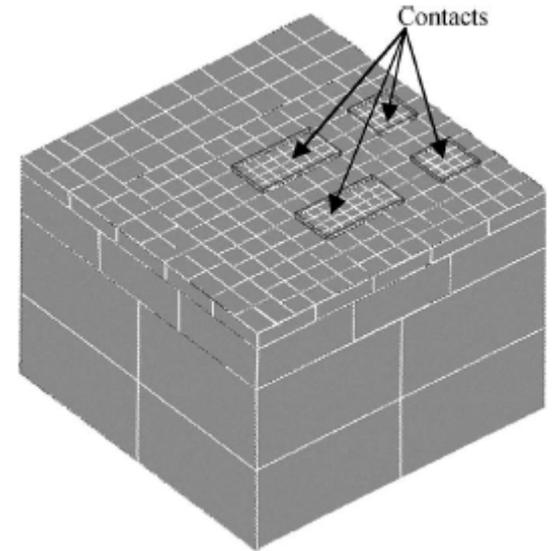


Consider both ohmic and displacement current;
Each substrate material has a conductivity σ_i , and a permittivity ϵ_i .

Z denotes the coupling impedance between the contacts (including the plane).

Direct Boundary Element Method (DBEM)

- Discretizes substrate boundary,
 - Contact surface
 - Medium boundary/interface
- Uses the simple free-space Green's function
- Has no difficulty in handling non-stratified substrates
- Efficient techniques of unknown reduction and matrix sparsification have been proposed for substrate resistance extraction [TCAD'06]



X. Wang, W. Yu and Z. Wang, "Efficient direct boundary element method for resistance extraction of substrate with arbitrary doping profiles," IEEE Trans. Comput. Aided Design, vol. 25, pp. 3035-3042, 2006.

High-Freq. Parameter Extraction using DBEM

- Within each medium

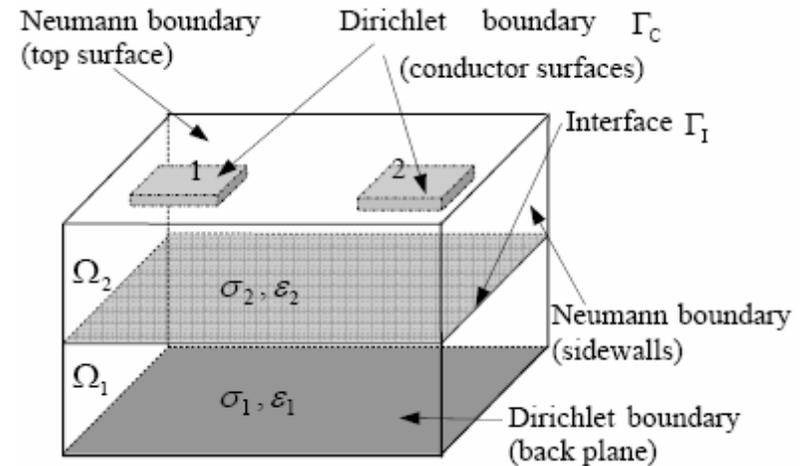
$$\nabla^2 u = \frac{\partial^2 u}{\partial x^2} + \frac{\partial^2 u}{\partial y^2} + \frac{\partial^2 u}{\partial z^2} = 0.$$

- Apply the Green's identities, with the weighting function of free-space Green's function

$$c_s u_s + \int_{\Gamma_i} q^* u d\Gamma = \int_{\Gamma_i} u^* q d\Gamma$$

u is electric potential, and *q* is its normal derivative on boundary

- Get linear equations with unknowns of *u* and *q* on boundaries
- *u* or *q* has known values on the outer boundaries



s is a collocation point;
u^{*} is the free-space Green's function, and
q^{*} is its normal derivative on boundary

High-Freq. Parameter Extraction using DBEM

- Along the interface of medium a and b:

$$u_a = u_b,$$

$$(\sigma_a + j\omega\varepsilon_a)E_{n,a} = (\sigma_b + j\omega\varepsilon_b)E_{n,b}.$$

For displacement current,
reflects the high-freq. model

- The linear equations for each medium can be combined together, to get an overall linear system

$$\mathbf{AX} = \mathbf{B}$$

A frequency-dependent complex-value system

To get whole impedance matrix, bias voltages are set on contacts, and \mathbf{B} reflects these settings. \mathbf{X} includes the unknowns of u and q on elements.

- The current flowing through contact k is

$$\int_{\Gamma_k} (\sigma + j\omega\varepsilon)E_n d\Gamma = \frac{1}{Z_{mk}} \quad \text{,if contact m is with 1V, and contact k 0V}$$

Outline

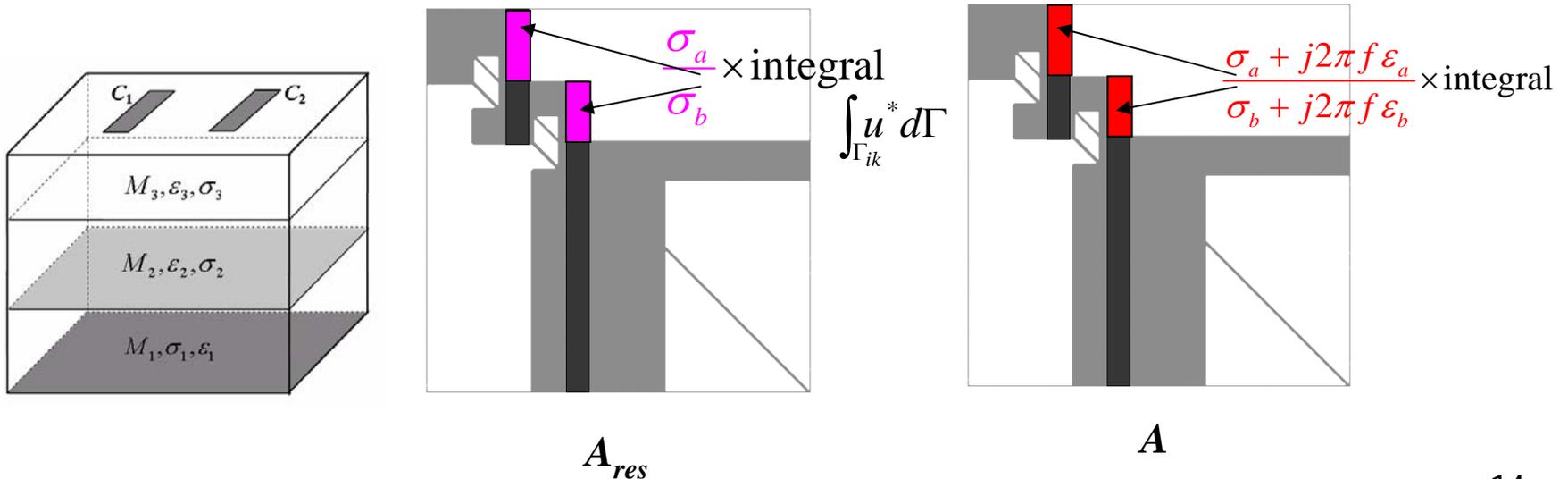
- Introduction of Substrate Coupling Problems
- High-Frequency Parameter Extraction Using DBEM
- **Efficient Techniques for Multi-Frequency Extraction**
 - **Extract substrate resistance (freq=0)**
 - **Perform one time**
 - **Revise solution of R extraction to freq.-dep. parameters**
 - **With improving techniques, easy for computation**
- Numerical Results
- Conclusions

Multi-Frequency Extraction of Substrate

- When consider both ohmic and displacement current, the substrate coupling parameter (Z) is frequency-dependent
- In DBEM calculation
 - Interface equation includes frequency and imaginary
$$(\sigma_a + j\omega\varepsilon_a)E_{n,a} = (\sigma_b + j\omega\varepsilon_b)E_{n,b}$$
 - Solve a frequency-dependent, complex-valued $\mathbf{AX} = \mathbf{B}$
 - Get freq.-dependent R and C from the complex-valued Z
- The substrate parameters for many frequencies are necessary for the knowledge of substrate coupling
- The trivial approach to build and solve $\mathbf{AX} = \mathbf{B}$ repeatedly loses its efficiency with the increase of frequency points
- We propose a fast method for the multi-frequency extraction

High Similarity of A_{res} and A

- Let A_{res} be the coefficient matrix for R extraction
- It's the equation for substrate resistance extraction, if we discard $j\omega\varepsilon$ items in $(\sigma_a + j\omega\varepsilon_a)E_{n,a} = (\sigma_b + j\omega\varepsilon_b)E_{n,b}$
- The differences between A_{res} and A lie on entries corresponding to q variables on interfaces



High Similarity of A_{res} and A (Cont.)

- In mathematics,

$$\mathbf{A} = \mathbf{A}_{res} + \mathbf{UV}^T$$

Where \mathbf{U} is a submatrix of \mathbf{A}_{res} , and \mathbf{V} is a sparse diagonal matrix.

The diagram illustrates the decomposition of the difference matrix $\mathbf{UV}^T = \mathbf{A} - \mathbf{A}_{res}$. It is shown as the product of a submatrix \mathbf{U} and a sparse diagonal matrix \mathbf{V}^T .

The first box represents \mathbf{UV}^T with two shaded rectangular blocks: $(f_{12}-r_{12})\mathbf{S}_{12}$ and $(f_{23}-r_{23})\mathbf{S}_{23}$.

The second box represents \mathbf{U} with two shaded rectangular blocks: $r_{12}\mathbf{S}_{12}$ and $r_{23}\mathbf{S}_{23}$.

The third box represents \mathbf{V}^T as a diagonal matrix with two shaded rectangular blocks: $(f_{12}-r_{12})/r_{12}\mathbf{I}_1$ and $(f_{23}-r_{23})/r_{23}\mathbf{I}_2$.

$$\mathbf{UV}^T = \mathbf{A} - \mathbf{A}_{res} = \mathbf{U} \times \mathbf{V}^T$$

$$f_{12} = (\sigma_2 + j\omega\epsilon_2)/(\sigma_1 + j\omega\epsilon_1), \quad r_{12} = \sigma_2/\sigma_1$$

\mathbf{S}_{12} , \mathbf{S}_{23} are matrices with entries calculated from the integrals of $\int_{\Gamma_{ik}} u^* d\Gamma$.
 \mathbf{I}_1 and \mathbf{I}_2 are identity matrices.

Revise R to Frequency-Dependent Z

- According to Sherman-Morrison-Woodbury formula,

$$\begin{aligned} A^{-1} &= (A_{res} + UV^T)^{-1} \\ &= A_{res}^{-1} - A_{res}^{-1}U(I + V^T A_{res}^{-1}U)^{-1}V^T A_{res}^{-1} \end{aligned}$$

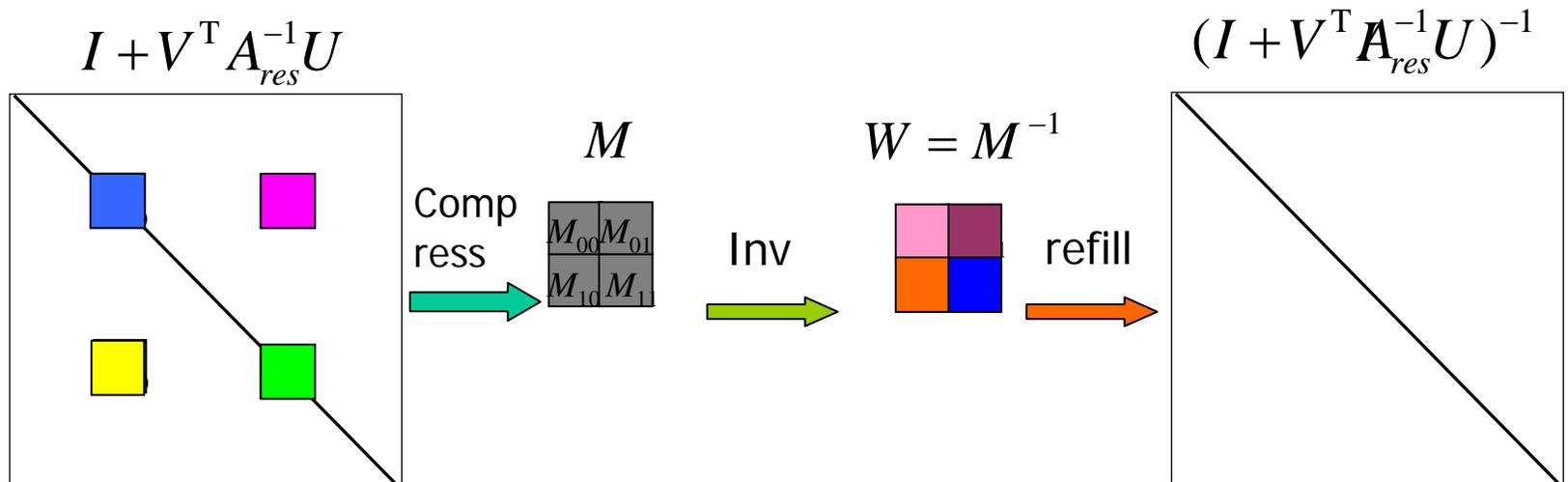
$$\begin{aligned} \Rightarrow X &= A^{-1}B = (A_{res} + UV^T)^{-1}B \\ &= A_{res}^{-1}B - A_{res}^{-1}U(I + V^T A_{res}^{-1}U)^{-1}V^T A_{res}^{-1}B \\ &= X_{res} - \{ A_{res}^{-1}U(I + V^T A_{res}^{-1}U)^{-1}V^T \} X_{res} \end{aligned}$$

- Thus, freq.-dept. solution (X) can be obtained through revising the solution for R extraction (X_{res})

Efficient Technique for $(I + V^T A_{res}^{-1} U)^{-1}$

$$X = X_{res} - A_{res}^{-1} U (I + V^T A_{res}^{-1} U)^{-1} V^T X_{res}$$

- Usually, difficult to get $(I + V^T A_{res}^{-1} U)^{-1}$
- U and V are sparse matrices, and so is $I + V^T A_{res}^{-1} U$
- For the three-medium example:



Algorithm Flow

1. Generate A_{res} in (5). Select some entries to form U ;
2. Solve for $X_{res} = A_{res}^{-1}B$ as well as for $A_{res}^{-1}U$;
3. For each frequency point, calculate the corresponding Z :
 - a) Create such factors as $(f_{12}/r_{12}-1)I_1$ and $(f_{23}/r_{23}-1)I_2$, so as to form matrix V ;
 - b) Compress $I + V^T A_{res}^{-1}U$ into small matrix M ;
 - c) Inverse matrix M , and get W ;
 - d) Refill W to get matrix $(I + V^T A_{res}^{-1}U)^{-1}$;
 - e) Get $X = X_{res} - A_{res}^{-1}U(I + V^T A_{res}^{-1}U)^{-1}V^T X_{res}$.
 - f) Get the desired Z parameter through (9).

Solve the real-valued problem of R extraction; Runs only once.

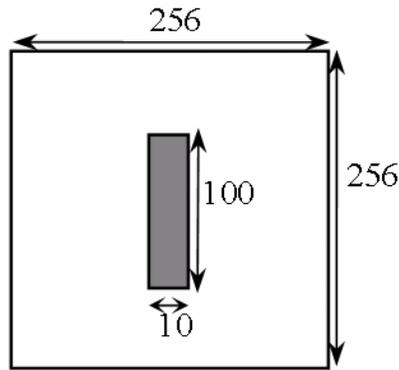
The order of M is # of interface elements, which is much less than total # of unknowns

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Numerical Results -- Accuracy

Case from [TCAD'98]



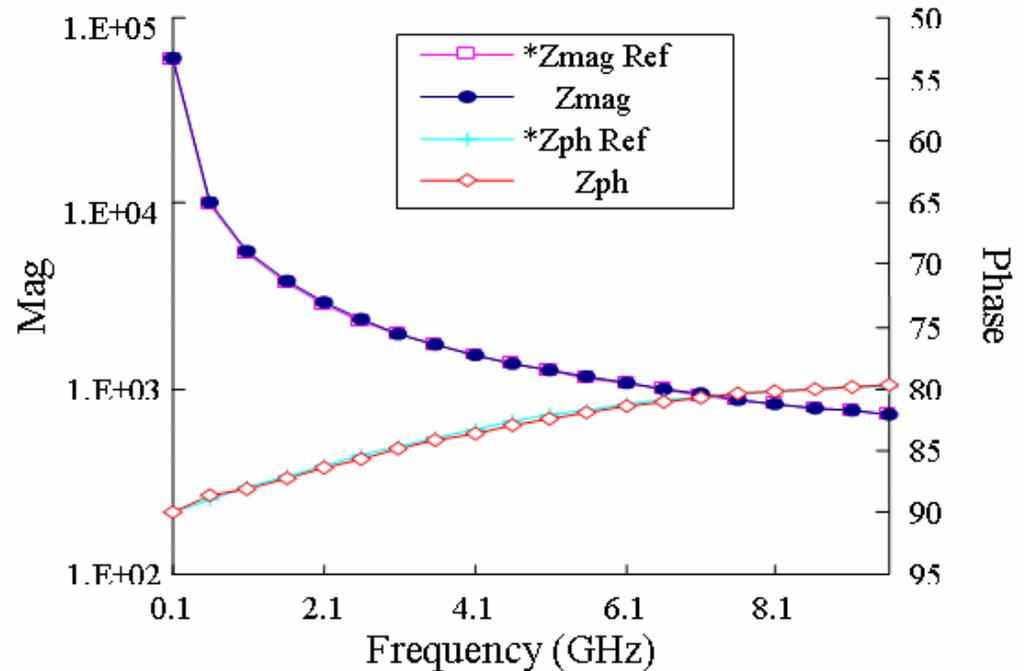
Top view

Contact

$\epsilon_{\text{ox}}=3.9$	$t_{\text{ox}}=1\mu\text{m}$
$\epsilon_{\text{ox}}=3.9$	$t_{\text{ox}}=1\mu\text{m}$
$\rho_{\text{epi}}=100\text{k}\Omega\text{-}\mu\text{m}$,	$t_{\text{epi}}=5\mu\text{m}$
$\rho_{\text{epi}}=100\Omega\text{-}\mu\text{m}$,	$t_{\text{sub}}=300\mu\text{m}$

Side view

The discrepancy between our results and those from [TCAD'98] < 1.0%



Since displacement current increases with frequency, the Z's magnitude decreases with frequency.

[TCAD'98] A.M. Niknejad, R. Gharpurey, and R.G. Meyer, "Numerically stable Green function for modeling and analysis of substrate coupling in integrated circuits," *IEEE Trans. CAD*, Vol. 17, pp. 305-315, 1998.

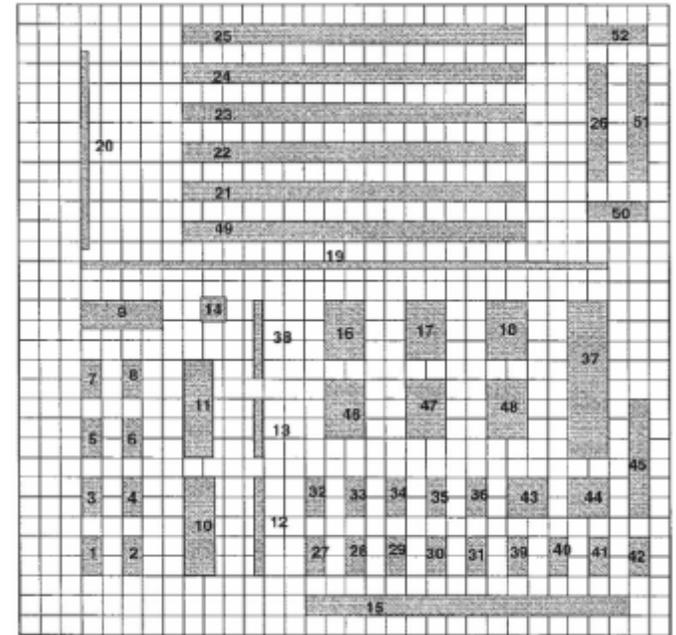
Numerical Results -- Efficiency

Case with 52 contacts

Our method is compared with ASITIC, a shared program using DCT-accelerated Green's function method

For each frequency, our method solves a 587×587 M matrix to calculate Z from the precalculated X_{res}

Our method is much faster than ASITIC



	DBEM		ASITIC
# Contacts	52		52
# Variables	7252	860	4352 (default)
Memory (MB)	60	32	310
Pre-process (seconds)	420.9*	360**	360 **
Extraction (for 1 freq.)	9.0	9000	(Not obtained)

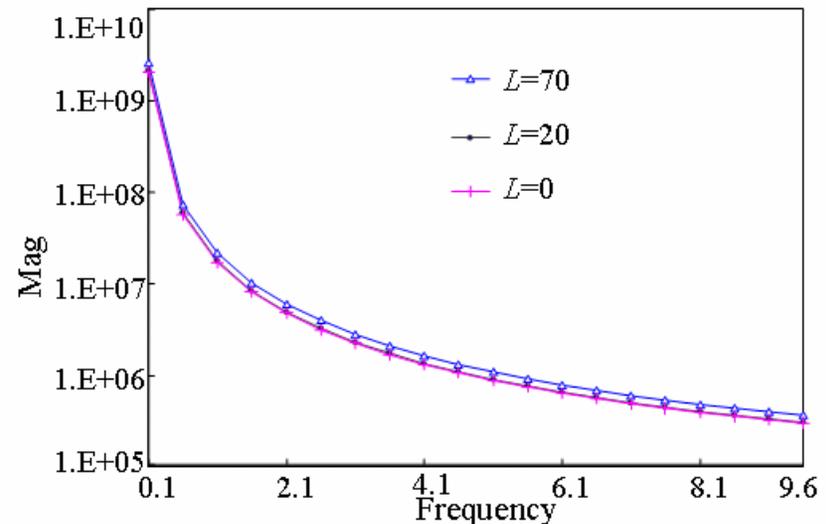
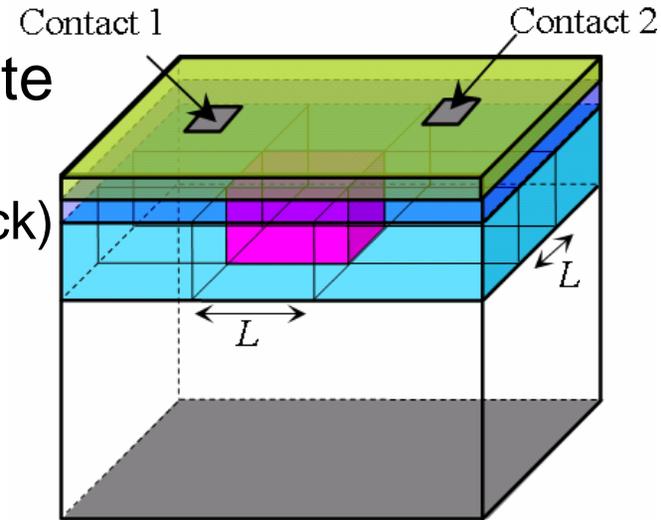
*: Calculate resistance

** : Calculate Green's function

Numerical Results -- Versatility

- An example with non-stratified substrate
 - The central block region has a distinct resistivity, called LVB (lateral variation block)
 - LVB's resistivity is 1000 times larger
 - Size of LVB is set to be 0, 20, and 70 μm
- Green's function based method is not able to handle it, while our DBEM has no difficulty
- Right figure shows the plot of Magnitude vs. frequency, for three settings

Since LVB obstructs ohmic current flow, the Z for $L=70$ is larger than those corresponding to other two settings



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Conclusions

- For frequency-dependent substrate modeling, we propose an efficient method for multi-frequency extraction
 - Based on DBEM, has ability to handle non-stratified structure
 - Do not directly solve the complex linear system, but firstly solve the R extraction problem with real linear system
 - For each frequency, the result of R extraction is revised to Z parameter through solving a smaller linear system
- The proposed method
 - Is efficient
 - Does not sacrifice accuracy
 - Has large versatility

Thank you!

For more information, please contact:

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