

Symbolic Model Checking of Analog/Mixed-Signal Circuits

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Motivation

- System on a chip (SoC) designs are becoming commonplace.
- They include digital and *analog/mixed-signal* (AMS) circuits.
- Validation is typically divided since digital requires long time steps while analog requires short time steps.
- Strict divisions in SoC validation, however, are rarely possible.
- New functional validation methods are needed to support the heterogeneous nature of SoC designs.

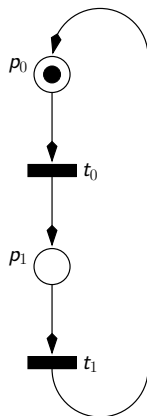
- Formal verification has shown advantages for digital circuits.
- Research has begun in formal methods for AMS circuits.
- AMS verification is complicated by the need to accurately track continuous quantities such as voltages and currents.

Background

- Alur et al. developed *hybrid automata* symbolic model-checking procedure for embedded systems.
- Seshia/Bryant developed timed automata symbolic model-checking procedure using BDDs.
- This work describes a hybrid Petri net symbolic model-checking procedure that uses BDDs for verifying analog/mixed-signal systems.
- Crucial to the acceptance of new AMS formal methods is the use of a familiar description language.
- The *labeled hybrid Petri net* (LHPN) model was developed to facilitate generation from a VHDL-AMS subset.
- We have developed a compiler that generates LHPNs given a circuit description using this subset of VHDL-AMS.

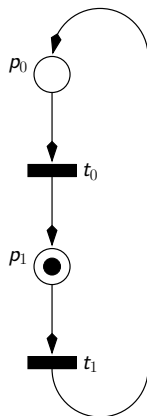
Labeled Hybrid Petri Nets (LHPNs)

- Composed of a Petri net and labels operating on continuous variables and Boolean signals.
- Label types are:
 - Enablings
 - Delay bounds
 - Boolean assignments
 - Value assignments
 - Rate assignments



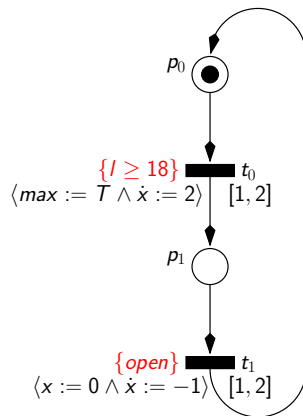
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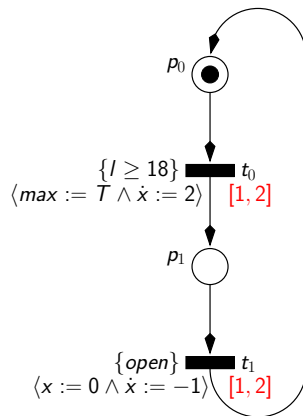
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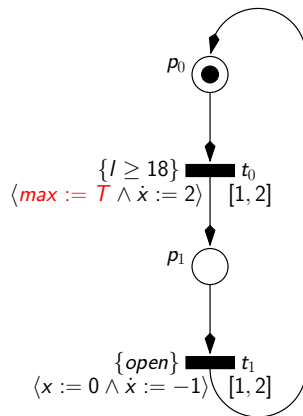
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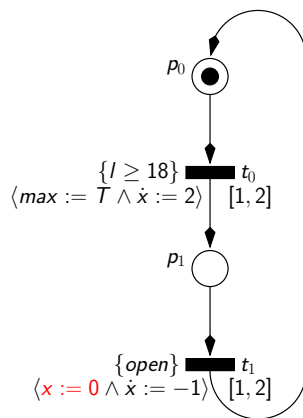
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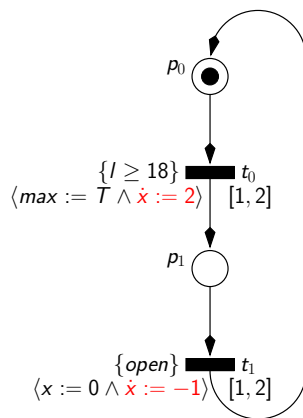
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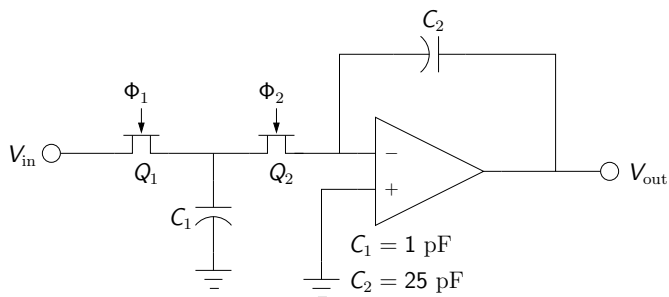


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Switched Capacitor Integrator Circuit

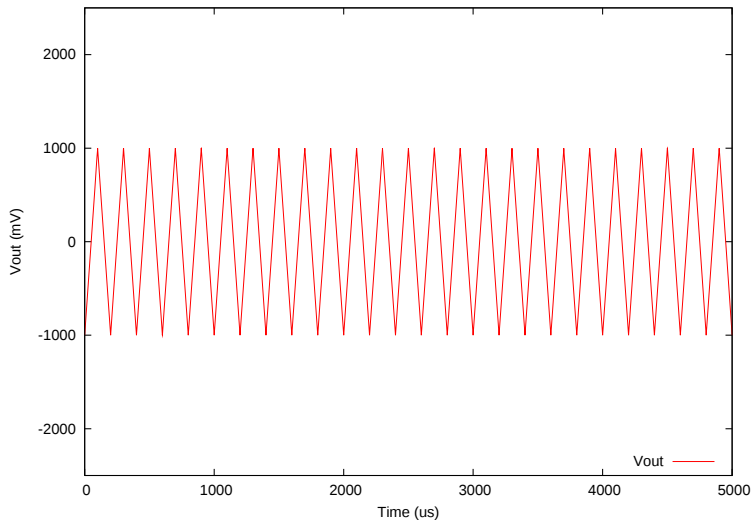


$$V_{\text{in}} = \pm 1000 \text{ mV}$$
$$\text{freq}(V_{\text{in}}) = 5 \text{ kHz}$$

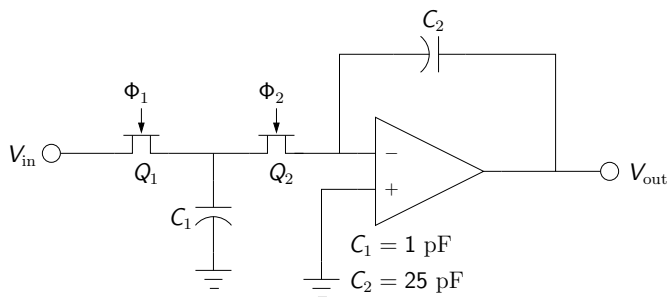
$$\text{freq}(\Phi_1) = \text{freq}(\Phi_2) = 500 \text{ kHz}$$
$$dV_{\text{out}}/dt = \pm 20 \text{ mV}/\mu\text{s}$$

Switched Capacitor Output Waveform

Typical Simulation



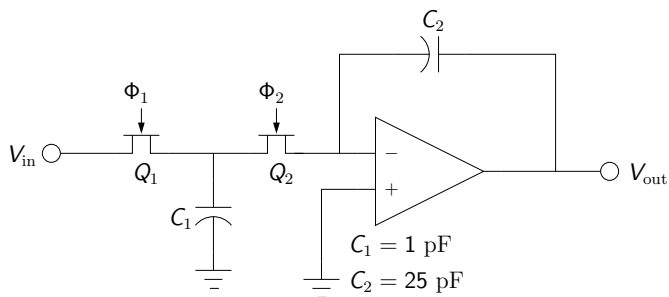
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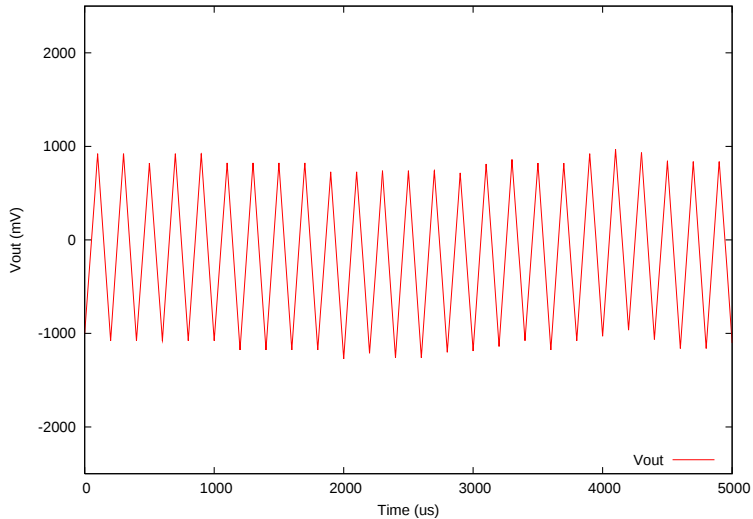
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Does V_{out} saturate? (i.e. $-2000\text{mV} \leq V_{out} \leq 2000\text{mV}$)

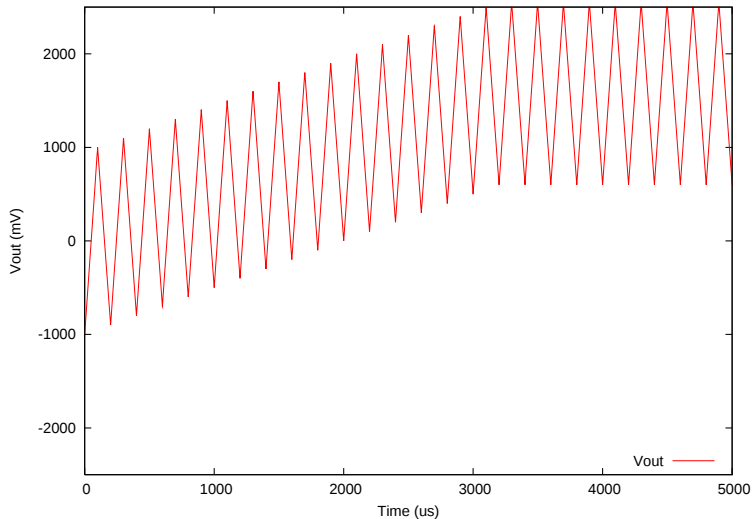
Switched Capacitor Output Waveform

Random Simulation



Switched Capacitor Output Waveform

Worst Case Simulation



Switched Capacitor Integrator Model

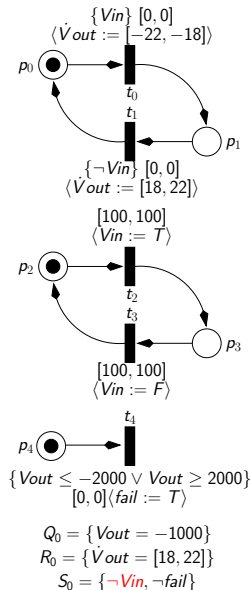
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use IEEE.std_logic_1164.all;
use work.handshake.all;
use work.nondeterminism.all;
entity integrator is
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architecture switchCap of integrator is
    signal Vin:std_logic := '0';
    quantity Vout:real;
begin
    break Vout => -1000.0; --Initial value
    if Vin='0' use
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    elsif Vin = '1' use
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    end use;
    process begin
        assign(Vin,'1',100,100);
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    end process;
    assert (Vout'above(-2000.0) and
        not Vout'above(2000.0))
        report "error"
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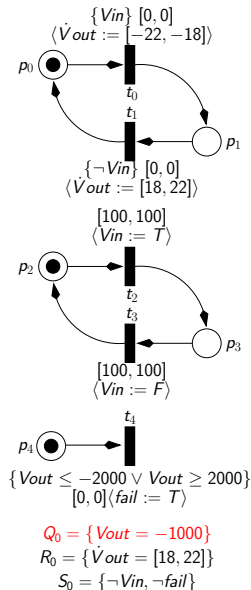


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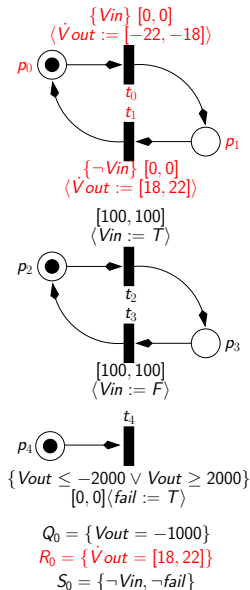


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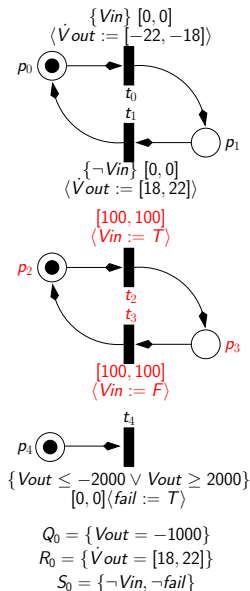


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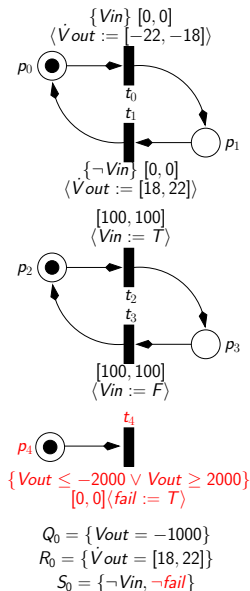


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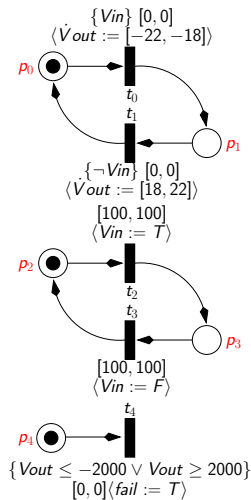
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Boolean Representation

Create Boolean variables for each of the following:

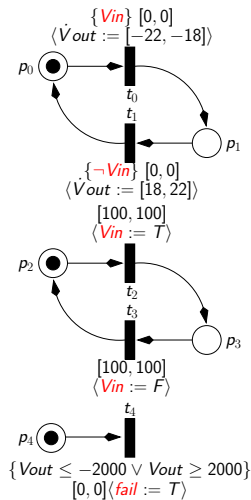
- Marking: $p_0, p_1, \dots, p_4$
- Boolean signals: $Vin, fail$
- Boolean rate variables:
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- Clock active variables:
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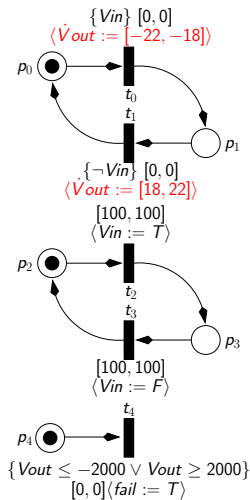
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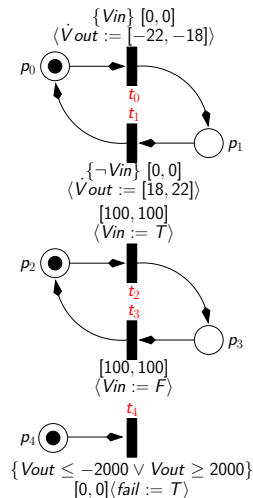


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Hybrid Separation Logic

- Maintain relationships among real values including system variables ($Vout$) and clock variables ($c_{t_0}, c_{t_1}, \dots, c_{t_4}$) using inequalities of the form $c_i x_i \geq c_j x_j + c$.
- Restrictions are placed upon inequalities to ensure canonicity.
- Boolean encoding:
 - Replace $c_i x_i \geq c_j x_j + c$ with a Boolean variable.
 - Note that $c_i x_i > c_j x_j + c$ is represented as $\overline{c_j x_j \geq c_i x_i + -c}$.
- Hybrid Separation Logic (HSL):

$$\phi ::= \mathbf{true} \mid \mathbf{false} \mid b \mid \neg\phi \mid \phi \wedge \phi \mid c_i x_i \geq c_j x_j + c$$

Boolean Constraints

Boolean constraints are periodically created as new inequalities are generated to maintain relationships among inequalities:

- Same variables, different constants:

$$2x \geq 3y + 5 \Rightarrow 2x \geq 3y + 4$$

- Same variables, same constants:

$$2x > 3y + 5 \Rightarrow 2x \geq 3y + 5$$

- Transitivity:

$$2x \geq 3y + 5 \wedge 4y \geq 5z + 5 \Rightarrow$$

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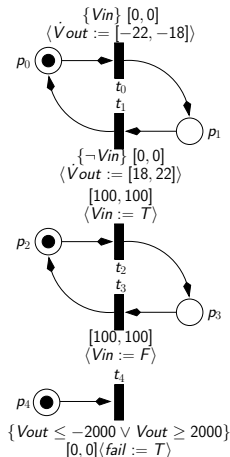
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Invariant

A statement that must always be satisfied.

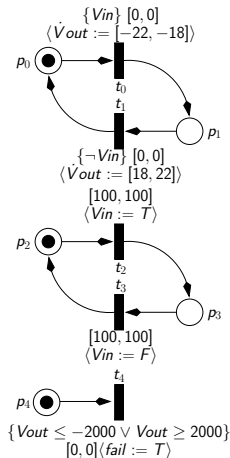
$$\phi_{\mathcal{I}} = \Phi \wedge \bigwedge_{t \in T} (a_t \Rightarrow \bullet t \wedge E(t) \wedge 0 \leq c_t \leq u(t)) \wedge (\overline{a_t} \Rightarrow \bullet \overline{t} \vee \widetilde{E(t)})$$



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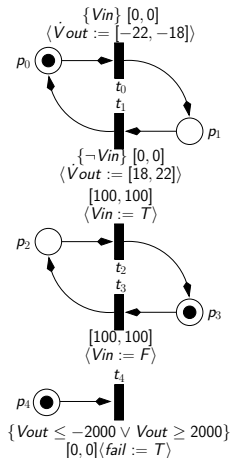


$$\begin{aligned} & p_0 \bar{p}_1 p_2 \bar{p}_3 p_4 \overline{Vin} \overline{fail} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \vee \\ & p_0 \bar{p}_1 \bar{p}_2 p_3 p_4 Vin \overline{fail} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \vee \\ & \bar{p}_0 p_1 \bar{p}_2 p_3 p_4 Vin \overline{fail} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \vee \\ & \bar{p}_0 p_1 p_2 \bar{p}_3 p_4 \overline{Vin} \overline{fail} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \end{aligned}$$

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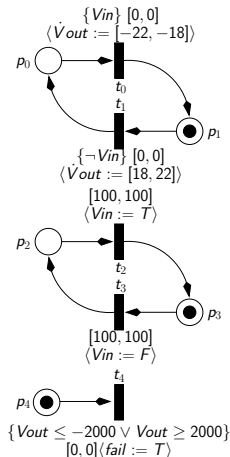
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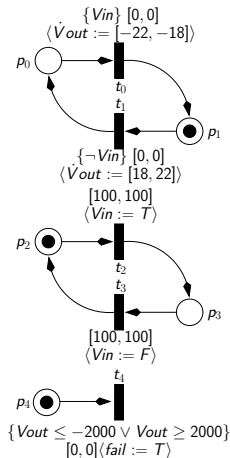
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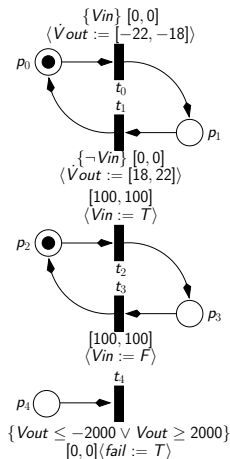


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Invariant

A statement that must always be satisfied.

$$\phi_{\mathcal{I}} = \Phi \wedge \bigwedge_{t \in T} (a_t \Rightarrow \bullet t \wedge E(t) \wedge 0 \leq c_t \leq u(t)) \wedge (\bar{a}_t \Rightarrow \bar{\bullet} t \vee \widetilde{E(t)})$$

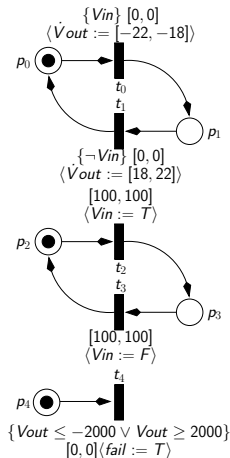


$$\begin{aligned}
 & p_0 \bar{p}_1 p_2 \bar{p}_3 p_4 \overline{Vin \text{ fail}} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \vee \\
 & p_0 \bar{p}_1 \bar{p}_2 p_3 p_4 Vin \overline{fail} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \vee \\
 & \bar{p}_0 p_1 \bar{p}_2 p_3 p_4 Vin \overline{fail} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \vee \\
 & \bar{p}_0 p_1 p_2 \bar{p}_3 p_4 \overline{Vin \text{ fail}} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \vee \\
 & p_0 \bar{p}_1 p_2 \bar{p}_3 \bar{p}_4 \overline{Vin \text{ fail}} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \vee \\
 & p_0 \bar{p}_1 \bar{p}_2 p_3 \bar{p}_4 Vin \overline{fail} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \vee \\
 & \bar{p}_0 p_1 \bar{p}_2 p_3 \bar{p}_4 Vin \overline{fail} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]} \vee \\
 & \bar{p}_0 p_1 p_2 \bar{p}_3 \bar{p}_4 \overline{Vin \text{ fail}} \overline{\dot{V}out_{[-22, -18]}} \dot{V}out_{[18, 22]}
 \end{aligned}$$

Invariant

A statement that must always be satisfied.

$$\phi_{\mathcal{I}} = \Phi \wedge \bigwedge_{t \in T} (a_t \Rightarrow \bullet t \wedge E(t) \wedge 0 \leq c_t \leq u(t)) \wedge (\bar{a}_t \Rightarrow \bullet \bar{t} \vee \widetilde{E(t)})$$



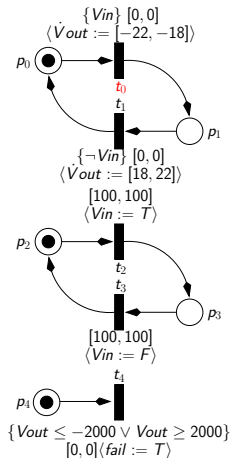
Pseudo-Inverse:

$$a \wedge \widetilde{x \leq 2000} = \bar{a} \vee x \geq 2000$$

Invariant

A statement that must always be satisfied.

$$\phi_{\mathcal{I}} = \Phi \wedge \bigwedge_{t \in T} (a_t \Rightarrow \bullet t \wedge E(t) \wedge 0 \leq c_t \leq u(t)) \wedge (\overline{a_t} \Rightarrow \overline{\bullet t} \vee \widetilde{E(t)})$$

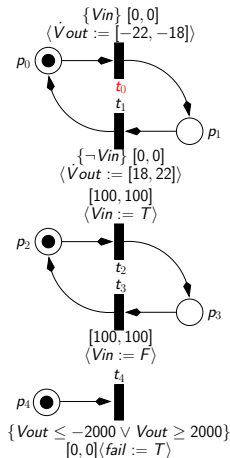


$$(a_{t_0} \Rightarrow p_0 \wedge Vin \wedge c_{t_0} = 0) \wedge (\overline{a_{t_0}} \Rightarrow \overline{p_0} \vee \overline{Vin})$$

Invariant

A statement that must always be satisfied.

$$\phi_{\mathcal{I}} = \Phi \wedge \bigwedge_{t \in T} (a_t \Rightarrow \bullet t \wedge E(t) \wedge 0 \leq c_t \leq u(t)) \wedge (\overline{a_t} \Rightarrow \overline{\bullet t} \vee \widetilde{E(t)})$$

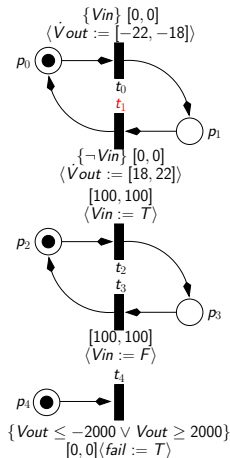


$$(a_{t_0} \Rightarrow p_0 \wedge Vin \wedge c_{t_0} = 0) \wedge (\overline{a_{t_0}} \Rightarrow \overline{p_0} \vee \overline{Vin})$$

Invariant

A statement that must always be satisfied.

$$\phi_{\mathcal{I}} = \Phi \wedge \bigwedge_{t \in T} (a_t \Rightarrow \bullet t \wedge E(t) \wedge 0 \leq c_t \leq u(t)) \wedge (\overline{a_t} \Rightarrow \overline{\bullet t} \vee \widetilde{E(t)})$$



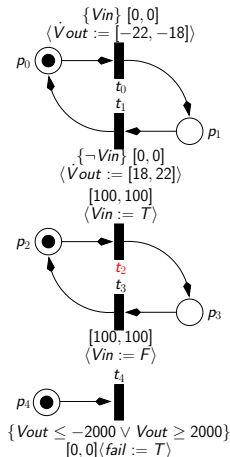
$$(a_{t_0} \Rightarrow p_0 \wedge Vin \wedge c_{t_0} = 0) \wedge (\overline{a_{t_0}} \Rightarrow \overline{p_0} \vee \overline{Vin}) \wedge$$

$$(a_{t_1} \Rightarrow p_1 \wedge \overline{Vin} \wedge c_{t_1} = 0) \wedge (\overline{a_{t_1}} \Rightarrow \overline{p_1} \vee Vin)$$

Invariant

A statement that must always be satisfied.

$$\phi_{\mathcal{I}} = \Phi \wedge \bigwedge_{t \in T} (a_t \Rightarrow \bullet t \wedge E(t) \wedge 0 \leq c_t \leq u(t)) \wedge (\overline{a_t} \Rightarrow \overline{\bullet t} \vee \widetilde{E(t)})$$

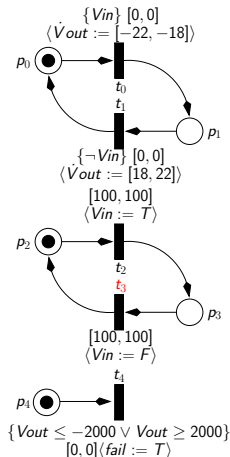


$$\begin{aligned} & (a_{t_0} \Rightarrow p_0 \wedge Vin \wedge c_{t_0} = 0) \wedge (\overline{a_{t_0}} \Rightarrow \overline{p_0} \vee \overline{Vin}) \wedge \\ & (a_{t_1} \Rightarrow p_1 \wedge \overline{Vin} \wedge c_{t_1} = 0) \wedge (\overline{a_{t_1}} \Rightarrow \overline{p_1} \vee Vin) \wedge \\ & (a_{t_2} \Rightarrow p_2 \wedge 0 \leq c_{t_2} \leq 100) \wedge (\overline{a_{t_2}} \Rightarrow \overline{p_2}) \end{aligned}$$

Invariant

A statement that must always be satisfied.

$$\phi_{\mathcal{I}} = \Phi \wedge \bigwedge_{t \in T} (a_t \Rightarrow \bullet t \wedge E(t) \wedge 0 \leq c_t \leq u(t)) \wedge (\overline{a_t} \Rightarrow \overline{\bullet t} \vee \widetilde{E(t)})$$

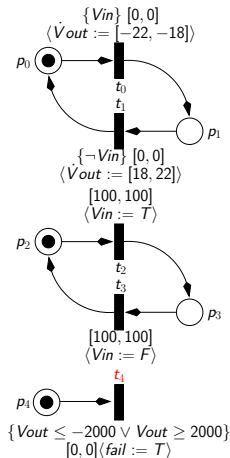


$$\begin{aligned} & (a_{t_0} \Rightarrow p_0 \wedge Vin \wedge c_{t_0} = 0) \wedge (\overline{a_{t_0}} \Rightarrow \overline{p_0} \vee \overline{Vin}) \wedge \\ & (a_{t_1} \Rightarrow p_1 \wedge \overline{Vin} \wedge c_{t_1} = 0) \wedge (\overline{a_{t_1}} \Rightarrow \overline{p_1} \vee Vin) \wedge \\ & (a_{t_2} \Rightarrow p_2 \wedge 0 \leq c_{t_2} \leq 100) \wedge (\overline{a_{t_2}} \Rightarrow \overline{p_2}) \wedge \\ & (a_{t_3} \Rightarrow p_3 \wedge 0 \leq c_{t_3} \leq 100) \wedge (\overline{a_{t_3}} \Rightarrow \overline{p_3}) \end{aligned}$$

Invariant

A statement that must always be satisfied.

$$\phi_{\mathcal{I}} = \Phi \wedge \bigwedge_{t \in T} (a_t \Rightarrow \bullet t \wedge E(t) \wedge 0 \leq c_t \leq u(t)) \wedge (\overline{a_t} \Rightarrow \overline{\bullet t} \vee \widetilde{E(t)})$$



$$\begin{aligned} & (a_{t_0} \Rightarrow p_0 \wedge Vin \wedge c_{t_0} = 0) \wedge (\overline{a_{t_0}} \Rightarrow \overline{p_0} \vee \overline{Vin}) \wedge \\ & (a_{t_1} \Rightarrow p_1 \wedge \overline{Vin} \wedge c_{t_1} = 0) \wedge (\overline{a_{t_1}} \Rightarrow \overline{p_1} \vee Vin) \wedge \\ & (a_{t_2} \Rightarrow p_2 \wedge 0 \leq c_{t_2} \leq 100) \wedge (\overline{a_{t_2}} \Rightarrow \overline{p_2}) \wedge \\ & (a_{t_3} \Rightarrow p_3 \wedge 0 \leq c_{t_3} \leq 100) \wedge (\overline{a_{t_3}} \Rightarrow \overline{p_3}) \wedge \\ & (a_{t_4} \Rightarrow p_4 \wedge c_{t_4} = 0 \wedge \\ & (Vout \leq -2000 \vee Vout \geq 2000)) \wedge \\ & (\overline{a_{t_4}} \Rightarrow \overline{p_4} \vee (Vout \geq -2000 \wedge Vout \leq 2000)) \end{aligned}$$

Primary Guarded Command Set

- Fires transition causing marking update and assignments.
- For each transition t_i :

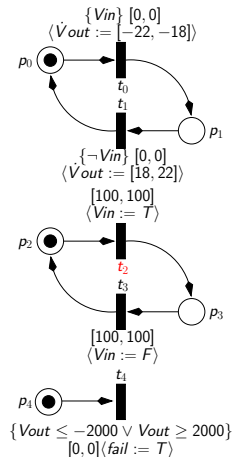
$$\mathcal{C}_P = \bigcup_{t \in T} \{ \langle \phi_{G_P}(t), \mathcal{A}_P(t) \rangle \}$$

where

$$\phi_{G_P}(t) = (\bullet t \wedge \overline{t \bullet - \bullet t} \wedge E(t) \wedge a_t \wedge c_t \geq l(t))$$

$$\mathcal{A}_P(t) = \{ (\bullet t - t \bullet) := F, (t \bullet) := T, a_t := F, c_t := [-\infty, \infty], \\ BA(t), VA(t), RA(t) \}$$

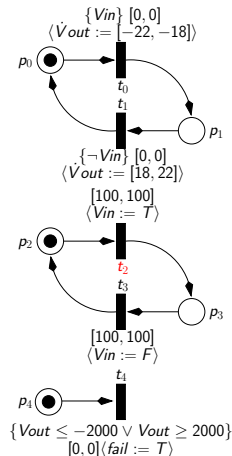
Primary Guarded Command for Integrator Example



$$\phi_{GP}(t_2) = p_2 \wedge \overline{p_3} \wedge a_{t_2} \wedge c_{t_2} \geq 100$$

$$\mathcal{A}_P(t_2) =$$

Primary Guarded Command for Integrator Example



$$\phi_{GP}(t_2) = p_2 \wedge \overline{p_3} \wedge a_{t_2} \wedge c_{t_2} \geq 100$$

$$\mathcal{A}_P(t_2) = \{p_2 := F, p_3 := T, a_{t_2} := F, \\ c_{t_2} := [-\infty, \infty], \\ Vin := T\}$$

Secondary Guarded Command Set

- Activates or deactivates clock based on marking, enabling condition, and clock value.
- For each transition t_i :

$$\mathcal{C}_S = \bigcup_{t \in T} \{ \langle \phi_{G_{SA}}(t), \mathcal{A}_{SA}(t) \rangle, \langle \phi_{G_{SD}}(t), \mathcal{A}_{SD}(t) \rangle \}$$

where

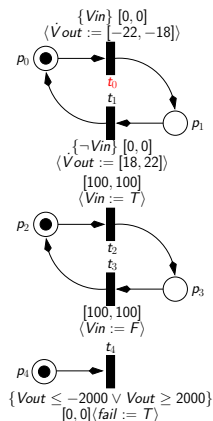
$$\phi_{G_{SA}}(t) = \bullet t \wedge E(t) \wedge \overline{a_t}$$

$$\mathcal{A}_{SA}(t) = \{a_t := T, c_t := [0, 0]\}$$

$$\phi_{G_{SD}}(t) = (\overline{\bullet t} \vee \widetilde{E(t)}) \wedge a_t$$

$$\mathcal{A}_{SD}(t) = \{a_t := F, c_t := [-\infty, \infty]\}$$

Secondary Guarded Commands for Integrator Example



Activating Guarded Command:

$$\phi_{G_{SA}}(t_0) = p_0 \wedge Vin \wedge \overline{a_{t_0}}$$

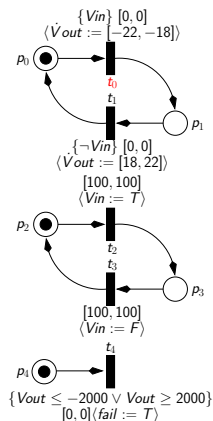
$$\mathcal{A}_{SA}(t_0) =$$

Deactivating Guarded Command:

$$\phi_{G_{SD}}(t_0) =$$

$$\mathcal{A}_{SD}(t_0) =$$

Secondary Guarded Commands for Integrator Example



Activating Guarded Command:

$$\phi_{GSA}(t_0) = p_0 \wedge Vin \wedge \overline{a_{t_0}}$$

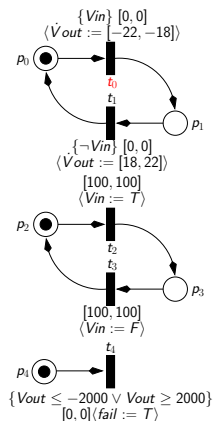
$$\mathcal{A}_{SA}(t_0) = \{a_{t_0} := T, c_{t_0} := [0, 0]\}$$

Deactivating Guarded Command:

$$\phi_{GSD}(t_0) =$$

$$\mathcal{A}_{SD}(t_0) =$$

Secondary Guarded Commands for Integrator Example



Activating Guarded Command:

$$\phi_{GSA}(t_0) = p_0 \wedge Vin \wedge \overline{a_{t_0}}$$

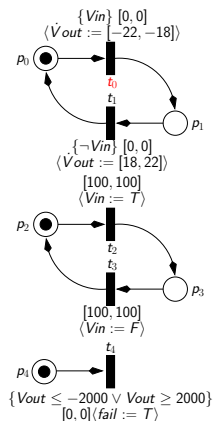
$$\mathcal{A}_{SA}(t_0) = \{a_{t_0} := T, c_{t_0} := [0, 0]\}$$

Deactivating Guarded Command:

$$\phi_{GSD}(t_0) = (\overline{p_0} \vee \overline{Vin}) \wedge a_{t_0}$$

$$\mathcal{A}_{SD}(t_0) =$$

Secondary Guarded Commands for Integrator Example



Activating Guarded Command:

$$\phi_{G_{SA}}(t_0) = p_0 \wedge Vin \wedge \overline{a_{t_0}}$$

$$\mathcal{A}_{SA}(t_0) = \{a_{t_0} := T, c_{t_0} := [0, 0]\}$$

Deactivating Guarded Command:

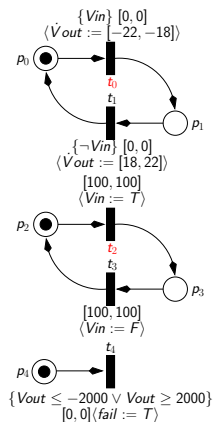
$$\phi_{G_{SD}}(t_0) = (\overline{p_0} \vee \overline{Vin}) \wedge a_{t_0}$$

$$\mathcal{A}_{SD}(t_0) = \{a_{t_0} := F, c_{t_0} := [-\infty, \infty]\}$$

Merging Primary and Secondary Guarded Commands

- Firing a transition may activate or deactivate clocks associated with other transitions, therefore primary and secondary guarded commands must be merged.
- Create new guarded commands when a primary assignment modifies a secondary guard.

Merging Guarded Commands for Integrator Example



$$\phi_G(t_2, t_0) = p_0 \wedge p_2 \wedge \overline{p_3} \wedge \overline{a_{t_0}} \wedge a_{t_2} \wedge c_{t_2} \geq 100$$

$$\mathcal{A}(t_2, t_0) = \{p_2 := F, p_3 := T, Vin := T, a_{t_0} := T, c_{t_0} := [0, 0], a_{t_2} := F, c_{t_2} := [-\infty, \infty]\}$$

Specifying Properties

- Dense real-time CTL is translated to a $\mathbf{T}\mu$ calculus formula.
- In the integrator example, the $\mathbf{T}\mu$ property is

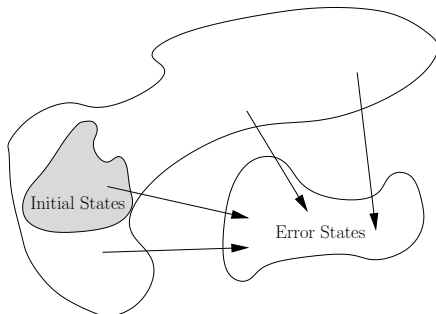
$$\phi_{init} \Longrightarrow \neg\mu Y.[fail \vee (\mathbf{true} \triangleright Y)]$$

where ϕ_{init} , the initial state, is:

$$\begin{aligned} \phi_{init} = & \overline{p_0} \overline{p_1} \overline{p_2} \overline{p_3} \overline{p_4} \overline{Vin} \overline{Fail} \overline{\dot{V}out_{[-22,-18]}} \overline{\dot{V}out_{[18,22]}} \\ & \overline{a_{t_0}} \overline{a_{t_1}} \overline{a_{t_2}} \overline{a_{t_3}} \overline{a_{t_4}} \wedge c_{t_2} = 0 \wedge Vout = -1000 \end{aligned}$$

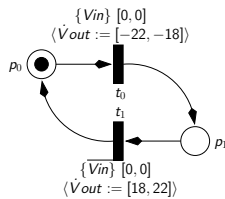
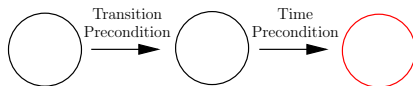
- If a state in which *fail* is true cannot be reached from the initial state, then the formula evaluates to true.

Model Checking Algorithm

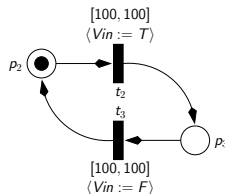


- Analysis proceeds by starting with a set of states that violate a property and finding all possible states that could have reached that state via time elapsing and/or firing transitions.
- If the initial state is encompassed within the calculated region, property is violated.

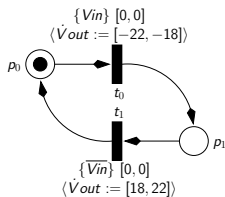
Time Precondition



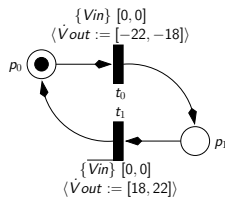
$Vout = 2000$
 $c_{t_2} = 100$



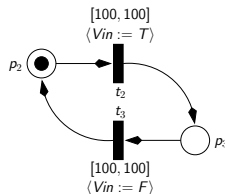
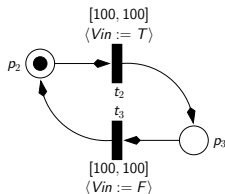
Time Precondition



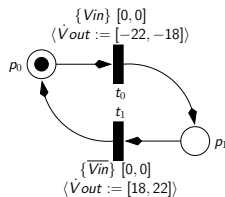
$$\begin{aligned}
 -200 &\leq Vout \leq 2000 \\
 0 &\leq c_{t_2} \leq 100 \\
 Vout &\geq 22c_{t_2} - 200 \\
 Vout &\leq 18c_{t_2} + 200
 \end{aligned}$$



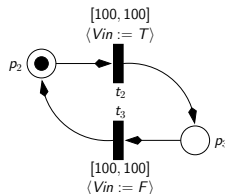
$$\begin{aligned}
 Vout &= 2000 \\
 c_{t_2} &= 100
 \end{aligned}$$



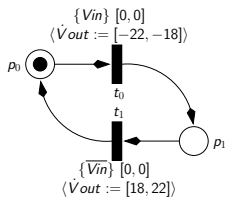
Transition Precondition



$$\begin{aligned} -200 &\leq Vout \leq 2000 \\ 0 &\leq c_{t_2} \leq 100 \\ Vout &\geq 22c_{t_2} - 200 \\ Vout &\leq 18c_{t_2} + 200 \end{aligned}$$

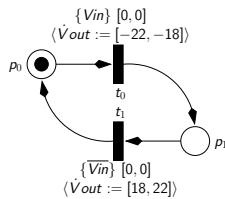


Transition Precondition



$$-200 \leq Vout \leq 200$$

$$c_{t_3} = 100$$

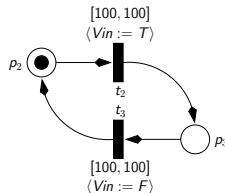
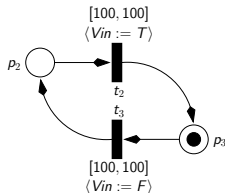


$$-200 \leq Vout \leq 2000$$

$$0 \leq c_{t_2} \leq 100$$

$$Vout \geq 22c_{t_2} - 200$$

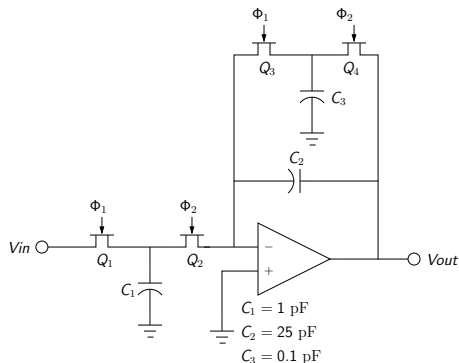
$$Vout \leq 18c_{t_2} + 200$$



Experimental Results

- Experimented with different ranges of rates for V_{out} in switched capacitor integrator circuit.
- With equal lower and upper rates, property is verified to be satisfied within a few seconds.
- With ranges of rates, property is verified to fail within a few seconds.

Corrected Switched Capacitor Integrator



$V_{in} = \pm 1 \text{ V}$
 $\text{freq}(V_{in}) = 5 \text{ kHz}$

$\text{freq}(\Phi_1) = \text{freq}(\Phi_2) = 500 \text{ kHz}$
 $dV_{out}/dt = \pm(16 \text{ to } 24) \text{ mV}/\mu\text{s}$

$$\frac{dV_{out}}{dt} = \begin{cases} 22 \text{ to } 24 \text{ mV}/\mu\text{s} & \text{when } V_{out} \leq -1000 \\ 16 \text{ to } 22 \text{ mV}/\mu\text{s} & \text{when } V_{out} > -1000 \\ -(22 \text{ to } 24) \text{ mV}/\mu\text{s} & \text{when } V_{out} > 1000 \\ -(16 \text{ to } 22) \text{ mV}/\mu\text{s} & \text{when } V_{out} \leq 1000 \end{cases}$$

Conclusions

- Hybrid system verification methods are necessary to verify analog/mixed-signal systems.
- It is possible to convert a subset of VHDL-AMS to LHPNs to encourage acceptance by AMS designers.
- Developed symbolic model checking method which uses BDDs for labeled hybrid Petri nets.
- Separation predicates are mapped to BDD variables.

- Provide a SPICE-deck front-end to further improve the ability of AMS designers to use our tool.
- Use abstraction methods to reduce number of BDD variables.
- Support handling of false negatives when applying abstraction.
- Implement algorithm using an SMT solver such as Yices.