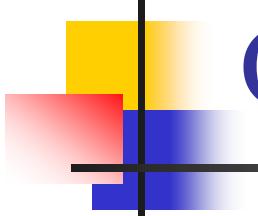


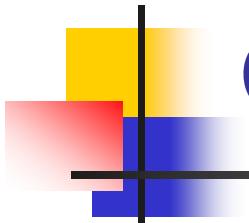
A Unified Multi-Corner Multi-Mode Static Timing Analysis Engine

Jing-Jia Nian, Shih-Heng Tsai,
Chung-Yang (Ric) Huang
National Taiwan University



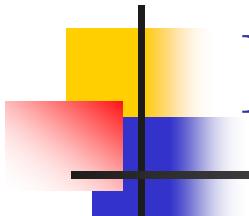
Outline

- Introduction & Problem formulation
- MCMM STA
- The Unified MCMM STA Engine
- Experimental Result
- Conclusion



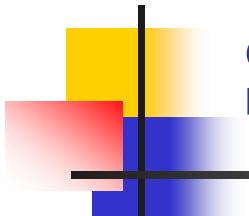
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Introduction

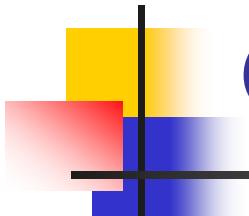
- Two approaches to conquer process variation in static timing analysis (STA)
 - Statistical STA
 - Models process variations as random variables
 - And models delay/transition time as PDFs
 - Corner-based STA
 - Focuses on worst/best cases only
 - Conservative but efficient



Statistical STA

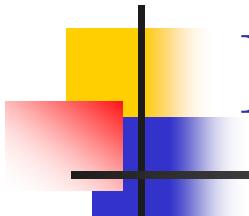
■ Defects of SSTA

- The variables may be asymmetric or uncertain and can not be modeled as Gaussian distribution
- Inabilities dealing with variable correlations
- The accuracy depends on the modeling provided by technology files which is not controllable by common users



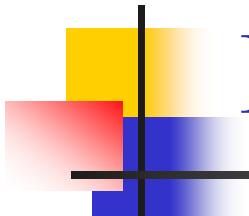
Corner-based STA

- Advantage
 - The timing model can be easily modified from the usual timing library
 - Linear time approach covering all corners exists
 - The same algorithm flow can be used to consider multiple “power modes”
- Disadvantage
 - The accumulated pessimistic boundary may cause the design hard to converge



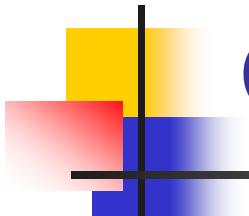
Proposed Techniques

- Modify the branch-and-bound method [Heloue, DATE07] to maintain upperbound quality
- Adopt linear time upper-bound technique [Onaissi, ICCAD06] to reduce the search space
- Devise an integrated parallel mechanic to achieve balance between performance and quality



Problem formulation

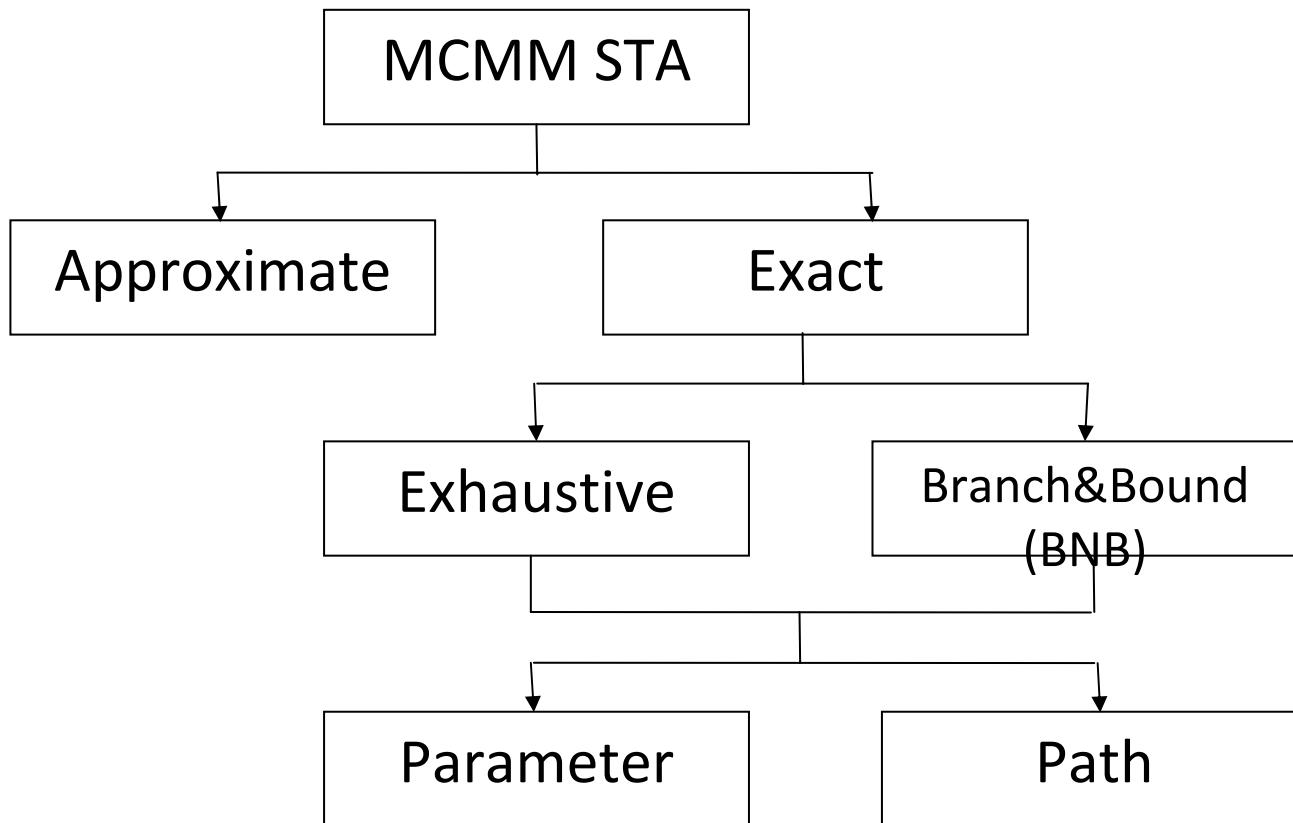
- Perform STA under multiple process parameters and only care about the corner values.
- Objective of this work:
 - To find the exact worst case delay/delay corner more efficiently and robustly than previous works.

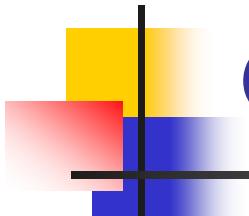


Outline

- Introduction & Problem formulation
- MCMM STA
 - Exact algorithms
 - Approximate algorithms
- The Unified MCMM STA Engine
- Experimental Result & Conclusion

Categorization of MCMM STA

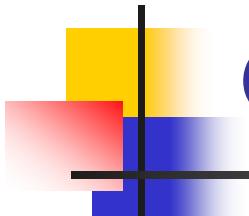




Complexity of Exact MCMM

- Given P parameters and N cells
- **Path-based** timing analysis
 - Add operation along a path is exact.
 - $(2-X_1+3X_2)+(6+2X_1-2X_2)=8+X_1+X_2$
 - For a single path, we can calculate all the timing corners together, which is linear to P.
 - However, the number of paths can grow exponentially with N, so the complexity is $O(P2^N)$.

Not favorable for circuits
with large number of Paths!!

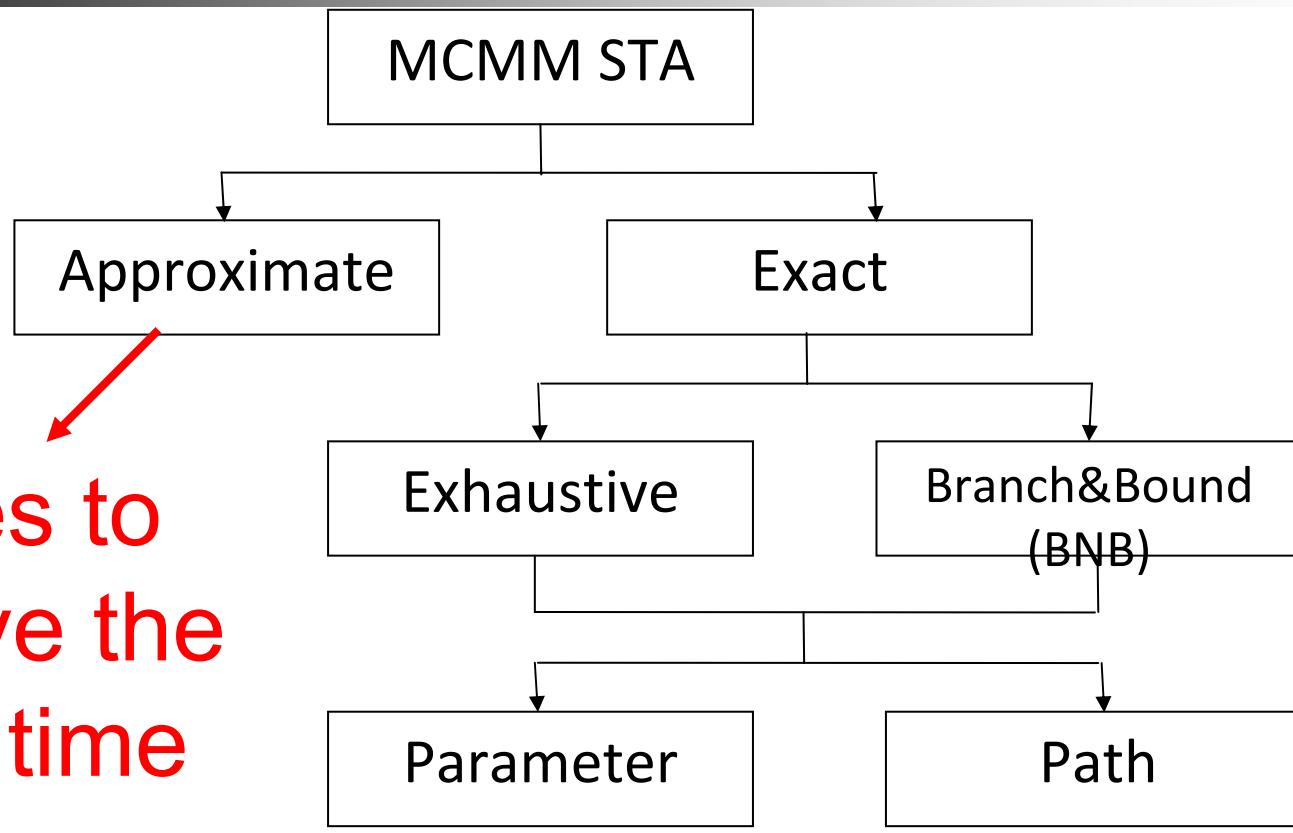


Complexity of Exact MCMM

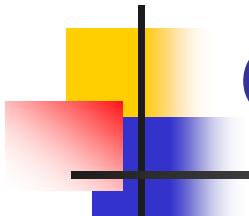
- **Parameter-based** timing analysis (**block-based**)
 - Max operation in a cell is not exact but an upper bound
 - $\max(2-X_1+3X_2, 6+2X_1-2X_2) = 6+2X_1+3X_2$
 - The complexity of the timing analysis for each corner is linear to N.
 - But we must run such a procedure for each of the 2^P corners, so the complexity is $O(N2^P)$.

Not favorable for large
number of corners!!!

Categorization of MCMM STA



Tries to
solve the
run time
issue

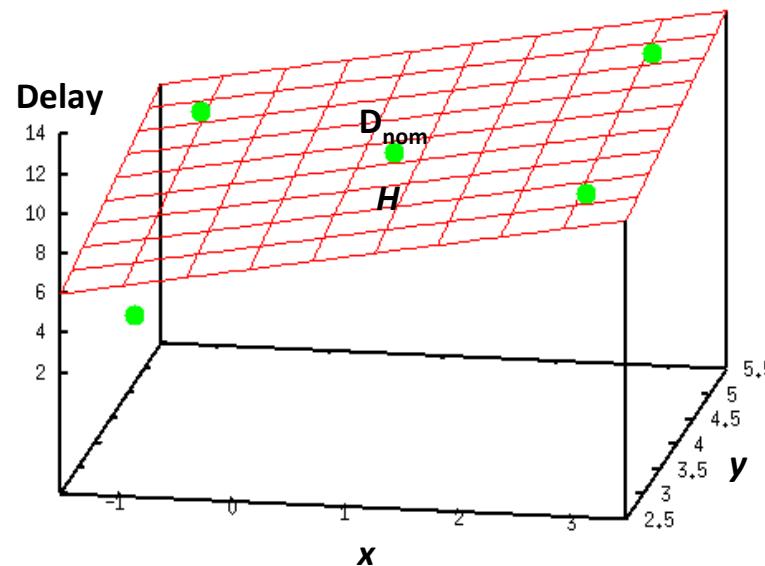


Complexity of Approximate MCMM

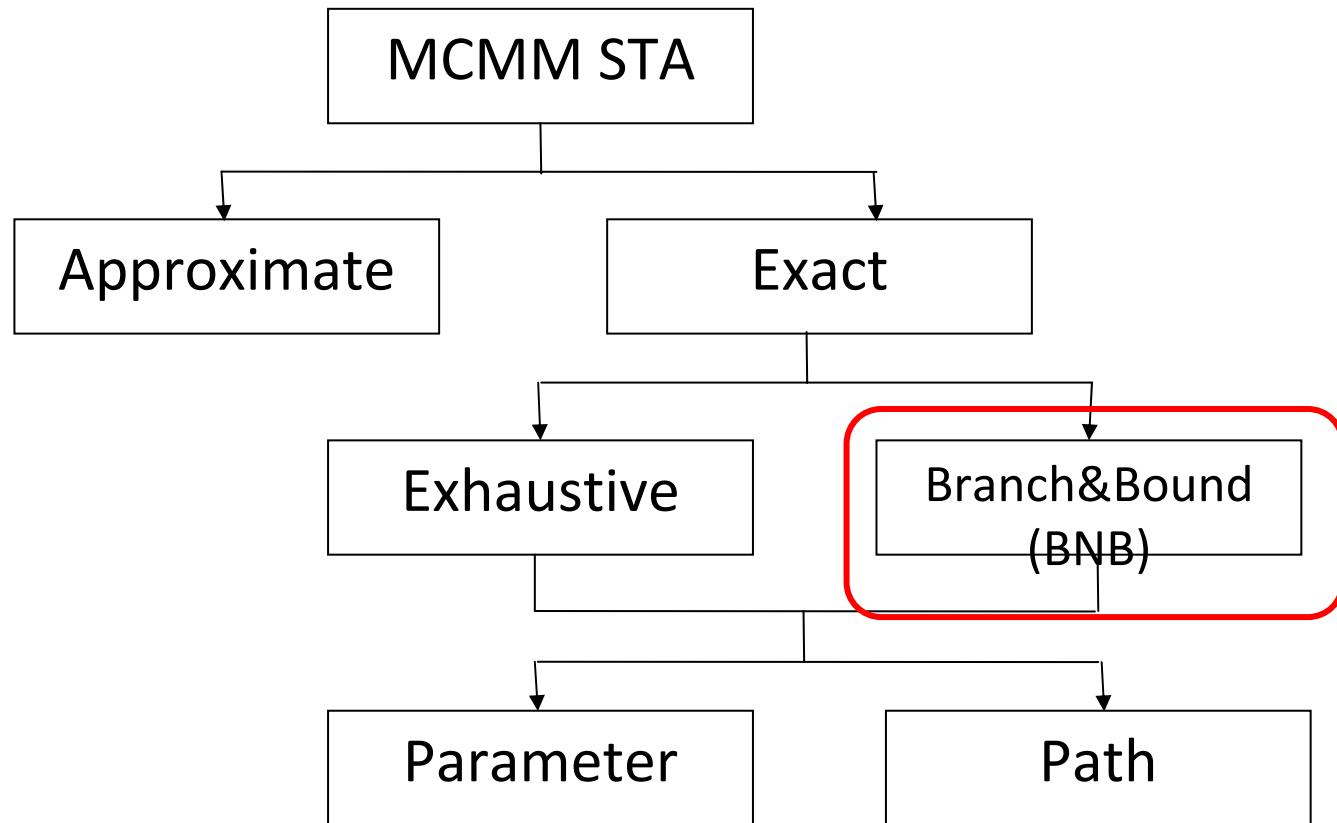
- Single-run block-based timing analysis
 - Performs max operation for N gates, each max operation costs $O(P)$.
 - Total complexity $O(NP)$, efficient but just an estimated upper bound.
 - “A Linear-Time approach for Static Timing Analysis Covering All Process Corners” in [Onaissi, ICCAD06]
- Different “max” operations:
 - Loose: choose the maximum coefficients.
 - $\text{maxLoose}(2-2X, 0+X) = 2+X$
 - Tight: not overestimate the overall peak value
 - $\text{maxTight}(2-2X, 0+X) = 2-X$

Linear Model of Process Variation

- Cell delays are modeled as linear functions of the process variations or operation modes.
 - $D = D_0 + a_1 X_1 + a_2 X_2 + \dots + a_{p-1} X_{p-1} + a_p X_p$

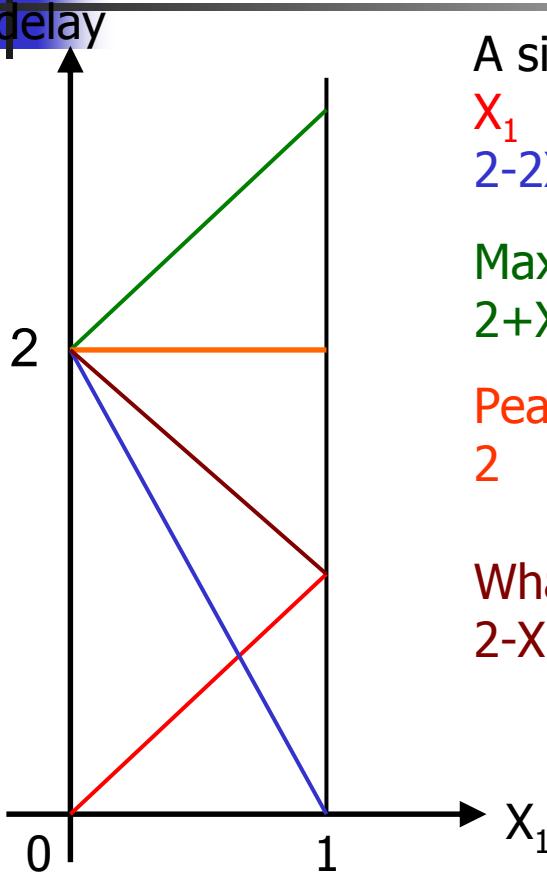


Categorization of MCMM STA



What if I want more accurate results?

The Tight and Loose Max Operation



A simple 1-D case:

$$x_1 \\ 2 - 2x_1$$

Max operation $O(P)$:

$$2 + x_1$$

Peak Value:
2

What we want $O(?)$:

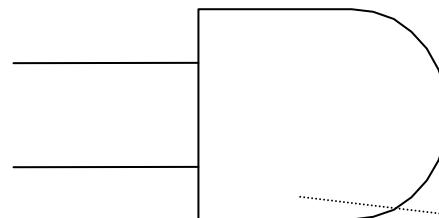
$$2 - x_1$$

- Finding the peak value of all corners and choose the highest $(P+1)$ points to form the upper bound. But the complexity is $O(2^P)$.
 - Linking $\langle(0), 2\rangle$ and $\langle(1), 1\rangle$ in this case
- $O(P)$ method to get tighter upper bound exists under the linear delay model.

Improvement on Pruning Power

$$D_{in1} = 3 + X_1 + X_2$$

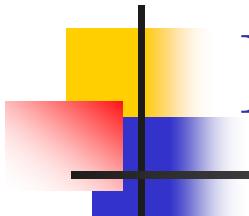
$$D_{in2} = 4 - X_1 - X_2$$



$$\begin{aligned} D_{out1} &= \textcolor{red}{maxLoose}(D_{in1}, D_{in2}) \\ &= 4 + X_1 + X_2, W_1 = \textcolor{blue}{6} \end{aligned}$$

$$\begin{aligned} D_{out2} &= \textcolor{red}{maxTight}(D_{in1}, D_{in2}) \\ &= 5 - 0.5X_1 - 0.5X_2, W_2 = \textcolor{blue}{5} \end{aligned}$$

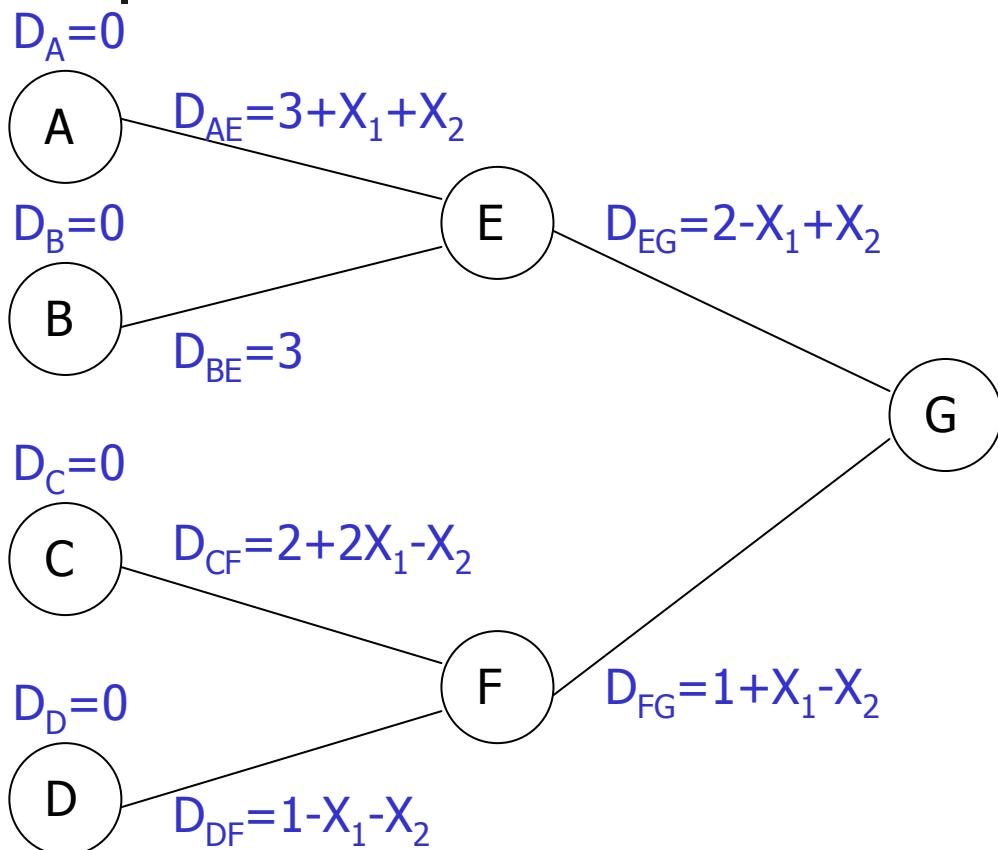
- If current worst delay is 5.5, then this group would be pruned with *maxTight* and would not be pruned with *maxLoose*.
- The difference could be used to improve pruning power!!



Branch-and-Bound MCMM STA

- Dynamic pruning methods (branch-and-bound)
 - Path-based BNB
 - Parameter-based BNB

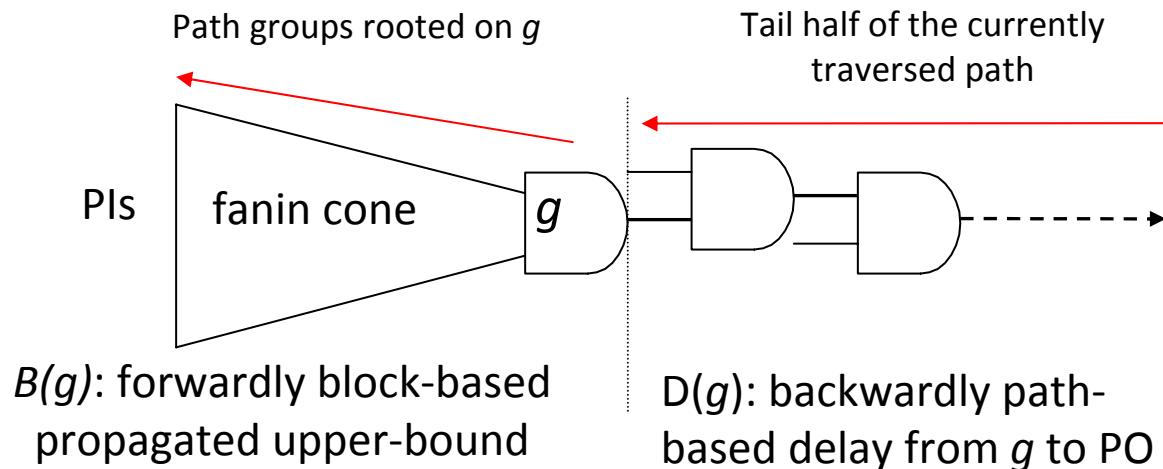
Path-based MCMM STA



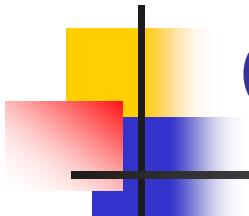
Path	Delay	Worst	Corner
AEG	$5 + 2X_2$	7	(X,1)
BEG	$5 - X_1 + X_2$	6	(0,1)
CFG	$3 + 3X_1 - 2X_2$	6	(1,0)
DFG	$2 - 2X_2$	2	(X,0)

Worst delay is 7 at the corner (X,1)

Path-based BNB



1. $D_{\text{current}} = \text{current maximum delay found}$
2. $(B(g) + D(g)) \leq D_{\text{current}}$, pruned.
3. if $(B(g) + D(g) > D_{\text{current}})$, continue to explore the fanin cone of g .

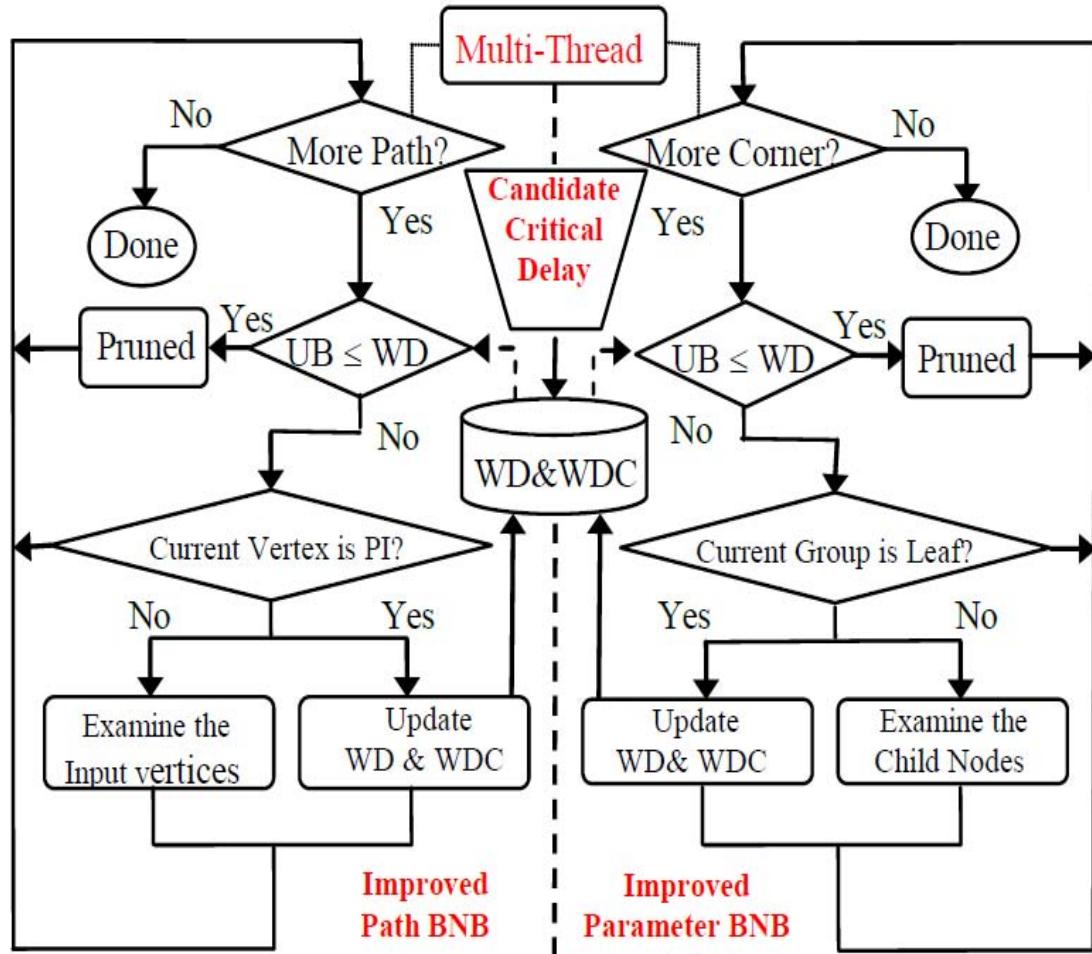


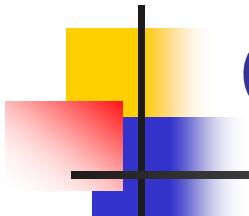
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Overview of the Unified Engine

- Improved BNB methods
- Multi-thread controller
- Candidate critical delay as initial bound.
- Share the worst delay/worst delay corner





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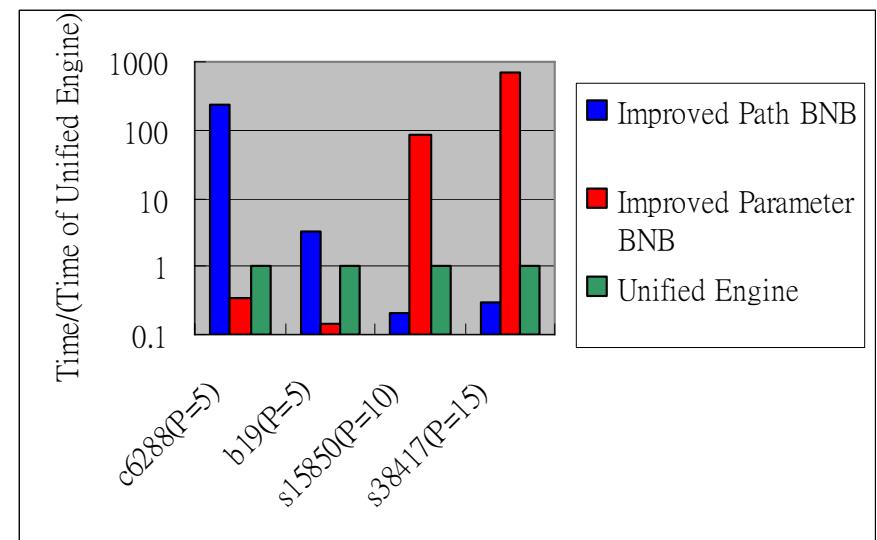
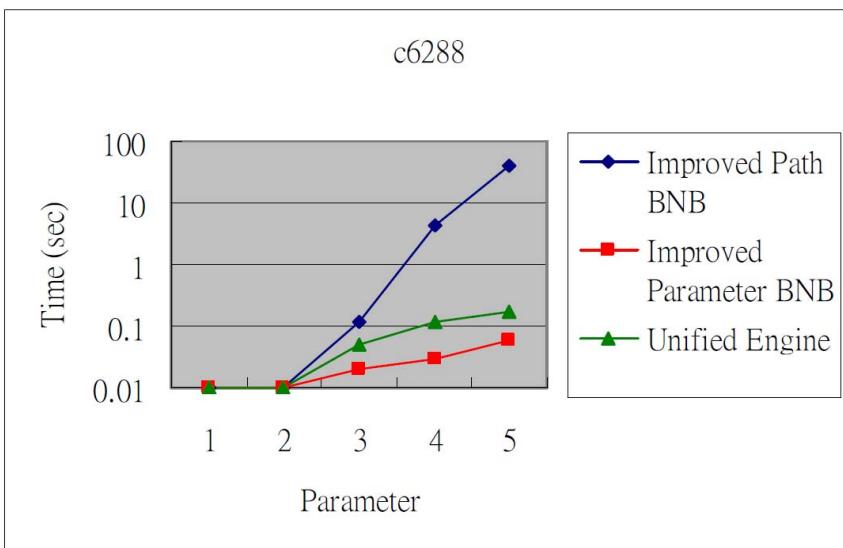
Improved Path BNB

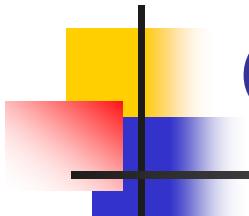
P=5	Circuit Information		Linear Time Approximate [6]		Exact computation (Path-based BNB)[9]		Our Exact computation (Path-based BNB)	
Circuit	Gate	Path	Delay (ns)	Run Time (s)	Delay (ns)	Run Time (s)	Delay (ns)	Run Time (s)
c6288	2.4k	10^{20}	42.2	0	N/A	-	39.9	40.4
s15850	9.7k	10^8	27.0	0.01	25.9	0.27	25.9	0.02
s38417	22k	10^6	15.1	0.02	14.6	0.02	14.6	0.02
s38584	19k	10^6	18.8	0.03	18.1	0.04	18.1	0.03
b18	111k	10^{24}	54.2	0.16	N/A	-	51.5	8.09
b19	224k	10^{25}	55.5	0.33	N/A	-	52.0	188

Improved Parameter BNB

c6288 gate counts: 2.4k		Exact computation (Parameter-based BNB)[9]		Our Exact computation (Parameter-based BNB)		
Parameter	Corners	Delay (ns)	Run Time (s)	Delay (ns)	Run Tim e (s)	Run Time Improveme nt
5	32	39.9	0.06	39.9	0.06	0%
6	64	40.2	0.13	40.2	0.13	0%
7	128	40.9	0.26	40.9	0.2	23%
8	256	40.8	0.5	40.8	0.38	24%
9	512	42.0	1.05	42.0	0.64	39%
10	1024	43.3	1.63	43.3	0.92	44%
15	32768	43.6	50.6	43.6	20.6	59%
20	1048576	44.6	1659	44.6	668	60%

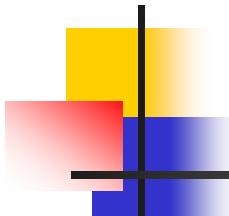
The Unified MCMM STA Engine





Conclusion

- We proposed a unified MCMM STA engine with:
 - A seamless integration of path and parameter BNB
 - An improved search space pruning technique
 - Candidate critical delay as initial bound
 - Extension to hold time check
- This robust engine will be a solid foundation for the MCMM TA research and a basis for MCMM timing optimization .



Reference

- [6] S. Onaissi and F. Najm, “A Linear-Time Approach for Static Timing Analysis Covering All Process Corners,” In Proceedings of IEEE Transactions on Computer-Aided Design of Integrated Circuits and Systems, vol. 27, 2008, page. 1291-1304.
- [7] K. Heloue and F. Najm, “Parameterized timing analysis with general delay models and arbitrary variation sources,” In Proceedings of Design Automation Conference, 2008, page. 403-408.
- [9] L. Silva, Miguel Silveira L z, and J. Phillips, “Efficient Computation of the Worst-Delay Corner,” In Proceedings of Design, automation and test in Europe, 2007, page. 1-6



Thank You & Have A Nice Day!

Q&A